#### 4: Linear Time Invariant

#### Systems

- LTI Systems
- Convolution Properties
- BIBO Stability
- Frequency Response
- Causality
- Convolution Complexity

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Linear Time-invariant (LTI) systems have two properties:

 $\text{Linear: } \mathscr{H}\left(\alpha u[n] + \beta v[n]\right) = \alpha \mathscr{H}\left(u[n]\right) + \beta \mathscr{H}\left(v[n]\right)$ 

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Convolution:  $x[n] * v[n] = \sum_{r=-\infty}^{\infty} x[r]v[n-r]$ 

Convolution obeys normal arithmetic rules for multiplication:

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Associative: x[n] \* (v[n] \* w[n]) = (x[n] \* v[n]) \* w[n] $\Rightarrow x[n] * v[n] * w[n]$  is unambiguous

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$$\begin{array}{l} \text{Distributive over +:} \\ x[n]*(\alpha v[n] + \beta w[n]) = (x[n]*\alpha v[n]) + (x[n]*\beta w[n]) \\ \text{Proof:} \ \sum_r x[n-r] \left(\alpha v[r] + \beta w[r]\right) = \\ \alpha \sum_r x[n-r]v[r] + \beta \sum_r x[n-r]w[r] \end{array}$$

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Distributive over +:  $x[n] * (\alpha v[n] + \beta w[n]) = (x[n] * \alpha v[n]) + (x[n] * \beta w[n])$ Proof:  $\sum_{r} x[n-r] (\alpha v[r] + \beta w[r]) =$   $\alpha \sum_{r} x[n-r]v[r] + \beta \sum_{r} x[n-r]w[r]$ 

Identity:  $x[n] * \delta[n] = x[n]$ 

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### BIBO Stability: Bounded Input, $x[n] \Rightarrow$ Bounded Output, y[n]

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 $\begin{array}{l} \operatorname{Proof}\left(1\right) \Rightarrow (2) \\ \operatorname{Define} x[n] = \begin{cases} 1 & h[-n] \geq 0 \\ -1 & h[-n] < 0 \\ & \operatorname{then} y[0] = \sum x[0-n]h[n] = \sum |h[n]|. \\ \operatorname{But}|x[n]| \leq 1 \forall n \text{ so BIBO} \Rightarrow y[0] = \sum |h[n]| < \infty. \end{array}$   $\begin{array}{l} \operatorname{Proof}\left(2\right) \Rightarrow (1) \\ \operatorname{Suppose} \sum |h[n]| = S < \infty \text{ and } |x[n]| \leq B \text{ is bounded.} \\ \operatorname{Then}|y[n]| = \left|\sum_{r=-\infty}^{\infty} x[n-r]h[r]\right| \end{array}$ 

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BIBO Stability: Bounded Input,  $x[n] \Rightarrow$  Bounded Output, y[n]

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Sign change in  $(1 + 2\cos\omega)$  at  $\omega = 2.1$  gives (a) gradient discontinuity in  $|H(e^{j\omega})|$ 

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Discontinuities of  $\pm k\pi$  do not affect group delay.



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Causal System: cannot see into the future i.e. output at time n depends only on inputs up to time n.

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#### Causal System: cannot see into the future

i.e. output at time n depends only on inputs up to time n.

### Formal definition:

If v[n] = x[n] for  $n \le n_0$  then  $\mathscr{H}(v[n]) = \mathscr{H}(x[n])$  for  $n \le n_0$ .

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Any right-sided sequence can be made causal by adding a delay. All the systems we will deal with are causal.

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$$y[n] = x[n] * h[n]$$
: convolve  $x[0: N-1]$  with  $h[0: M-1]$ 



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y[n] is only non-zero in the range  $0 \le n \le M+N-2$ 





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- y[n] is only non-zero in the range 0 < n < M + N 2
- Thus y[n] has only M + N 1 non-zero values



N = 8, M = 3M + N - 1 = 10

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$$x[n-r] \neq 0 \Rightarrow 0 \le n-r \le N-1$$





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#### Algebraically:

$$\begin{aligned} x[n-r] \neq 0 &\Rightarrow 0 \leq n-r \leq N-1 \\ &\Rightarrow n+1-N \leq r \leq n \end{aligned}$$



$$N = 8, M = 3$$
  
 $M + N - 1 = 10$ 

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Algebraically:

$$\begin{aligned} x[n-r] \neq 0 &\Rightarrow 0 \le n-r \le N-1 \\ &\Rightarrow n+1-N \le r \le n \end{aligned}$$
  
Hence:  $y[n] = \sum_{r=\max(0,n+1-N)}^{\min(M-1,n)} h[r]x[n-r]$ 



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$$x[n-r] \neq 0 \Rightarrow 0 \le n-r \le N-1 \\ \Rightarrow n+1-N \le r \le n$$

Hence: 
$$y[n] = \sum_{r=\max(0,n+1-N)}^{\min(M-1,n)} h[r]x[n-r]$$

We must multiply each h[n] by each x[n] and add them to a total  $\Rightarrow$  total arithmetic complexity (× or + operations)  $\approx 2MN$ 

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 $y_{\circledast}[n] = x[n] \circledast_N h[n]$ : circ convolve x[0:N-1] with h[0:M-1]



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N = 8, M = 3

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 $y_{\circledast_N}[n]$  has period N $\Rightarrow y_{\circledast_N}[n]$  has N distinct values



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- N = 8, M = 3
- Only the first M-1 values are affected by the circular repetition:

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$$N = 8, M = 3$$

- Only the first M-1 values are affected by the circular repetition:  $y_{\circledast_N}[n]=y[n]$  for  $M-1\leq n\leq N-1$
- If we append M 1 zeros (or more) onto x[n], then the circular repetition has no effect at all and:

 $y_{\circledast_{N+M-1}}[n] = y[n] \text{ for } 0 \le n \le N+M-2$ 

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Circular convolution is a necessary evil in exchange for using the DFT

# **Frequency-domain convolution**

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Idea: Use DFT to perform circular convolution - less computation

# **Frequency-domain convolution**

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Idea: Use DFT to perform circular convolution - less computation (1) Choose  $L \ge M + N - 1$  (normally round up to a power of 2)
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(1) Choose 
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(2) Zero pad x[n] and h[n] to give sequences of length L:  $\tilde{x}[n]$  and  $\tilde{h}[n]$ 

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Total operations:  $\approx 12L \log_2 L \approx 12 (M+N) \log_2 (M+N)$ 

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Beneficial if both M and N are  $> \sim 70$  .



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**Example:**  $M = 10^3$ ,  $N = 10^4$ :



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Example:  $M = 10^3$ ,  $N = 10^4$ : Direct:  $2MN = 2 \times 10^7$ with DFT:  $= 1.8 \times 10^6 \odot$ 



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Example:  $M = 10^3$ ,  $N = 10^4$ : Direct:  $2MN = 2 \times 10^7$ with DFT:  $= 1.8 \times 10^6 \odot$ 

But: (a) DFT may be very long if N is large (b) No outputs until all x[n] has been input.



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#### If N is very large:

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If N is very large: (1) chop x[n] into  $\frac{N}{K}$  chunks of length K



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If N is very large: (1) chop x[n] into  $\frac{N}{K}$  chunks of length K(2) convolve each chunk with h[n]



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Each output chunk is of length K + M - 1 and overlaps the next chunk

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If N is very large: (1) chop x[n] into  $\frac{N}{K}$  chunks of length K(2) convolve each chunk with h[n](3) add up the results



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Example: 
$$M = 500$$
,  $K = 10^4$ ,  $N = 10^7$   
Direct:  $2MN = 10^{10}$ 

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$$M = 500, K = 10^4, N = 10^7$$
  
Direct:  $2MN = 10^{10}$   
single DFT:  $12(M + N) \log_2(M + N) = 2.8 \times 10^9$ 

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Each output chunk is of length K + M - 1 and overlaps the next chunk Operations:  $\approx \frac{N}{K} \times 8 (M + K) \log_2 (M + K)$ Computational saving if  $\approx 100 < M \ll K \ll N$ 

Example:  $M = 500, K = 10^4, N = 10^7$ Direct:  $2MN = 10^{10}$ single DFT:  $12 (M + N) \log_2 (M + N) = 2.8 \times 10^9$ overlap-add:  $\frac{N}{K} \times 8 (M + K) \log_2 (M + K) = 1.1 \times 10^9$   $\odot$ 

Other advantages:

(a) Shorter DFT

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```

Other advantages:

- (a) Shorter DFT
- (b) Can cope with  $N=\infty$

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If N is very large: (1) chop x[n] into  $\frac{N}{K}$  chunks of length K(2) convolve each chunk with h[n](3) add up the results



Each output chunk is of length K + M - 1 and overlaps the next chunk Operations:  $\approx \frac{N}{K} \times 8 (M + K) \log_2 (M + K)$ Computational saving if  $\approx 100 < M \ll K \ll N$ 

```
Example: M = 500, K = 10^4, N = 10^7

Direct: 2MN = 10^{10}

single DFT: 12 (M + N) \log_2 (M + N) = 2.8 \times 10^9

overlap-add: \frac{N}{K} \times 8 (M + K) \log_2 (M + K) = 1.1 \times 10^9 \odot
```

Other advantages:

- (a) Shorter DFT
- (b) Can cope with  $N=\infty$
- (c) Can calculate y[0] as soon as x[K-1] has been read: algorithmic delay = K-1 samples

4: Linear Time Invariant Systems

- LTI Systems
- Convolution Properties
- BIBO Stability
- Frequency Response
- Causality
- Convolution Complexity

+

- Circular Convolution
- Frequency-domain

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#### Alternative method:

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Alternative method: (1) chop x[n] into  $\frac{N}{K}$  overlapping chunks of length K + M - 1



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Alternative method: (1) chop x[n] into  $\frac{N}{K}$  overlapping chunks of length K + M - 1(2)  $\circledast_{K+M-1}$  each chunk with h[n]



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The first M-1 points of each output chunk are invalid

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Alternative method: (1) chop x[n] into  $\frac{N}{K}$  overlapping chunks of length K + M - 1(2)  $\circledast_{K+M-1}$  each chunk with h[n](3) discard first M - 1 from each chunk



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The first M-1 points of each output chunk are invalid

Operations: slightly less than overlap-add because no addition needed to create y[n]

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The first M-1 points of each output chunk are invalid

Operations: slightly less than overlap-add because no addition needed to create y[n]

Advantages: same as overlap add

Strangely, rather less popular than overlap-add

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• LTI systems: impulse response, frequency response, group delay

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  - single DFT
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#### For further details see Mitra: 4 & 5.
## **MATLAB routines**

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fftfilt	Convolution using overlap add
$x[n] \circledast y[n]$	real(ifft(fft(x).*fft(y)))