

5: Filters

- Difference Equations
- FIR Filters
- FIR Symmetries +
- IIR Frequency Response
- Negating z +
- Cubing z +
- Scaling z +
- Low-pass filter +
- Allpass filters +
- Group Delay +
- Minimum Phase +
- Linear Phase Filters
- Summary
- MATLAB routines

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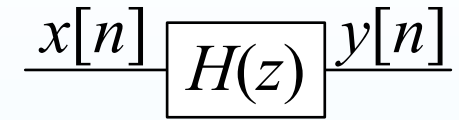
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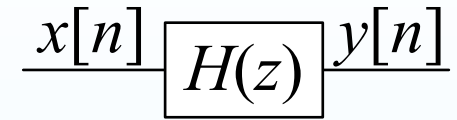
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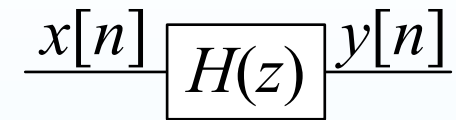
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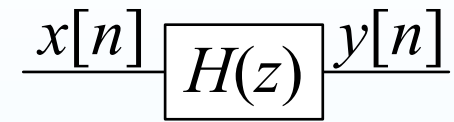
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Note **negative sign** in first equation.

Authors in some SP fields reverse the sign of the $a[n]$: **BAD IDEA**.

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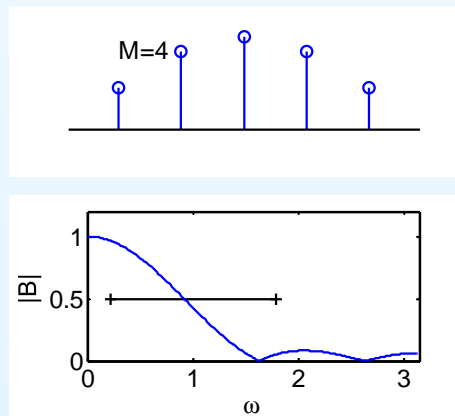
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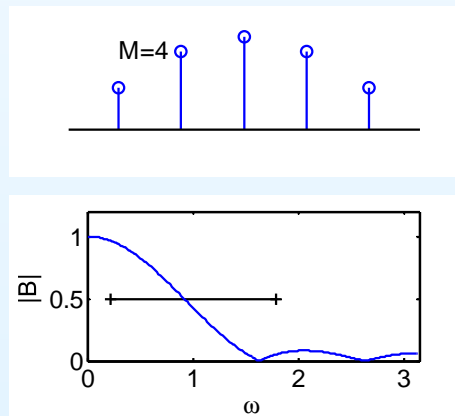
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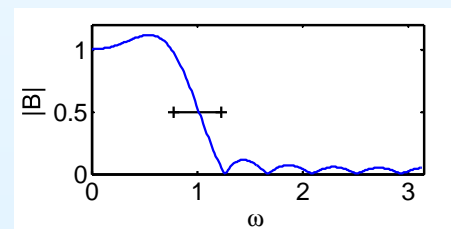
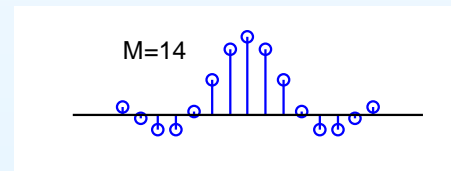
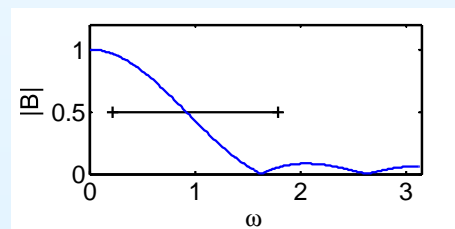
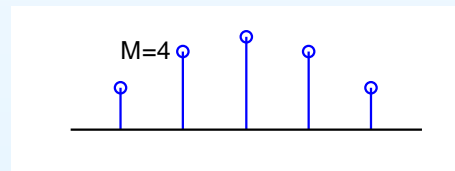
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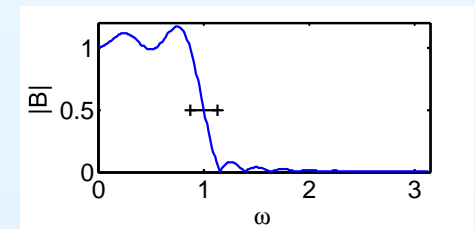
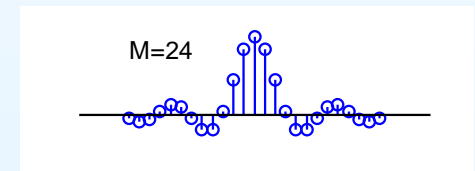
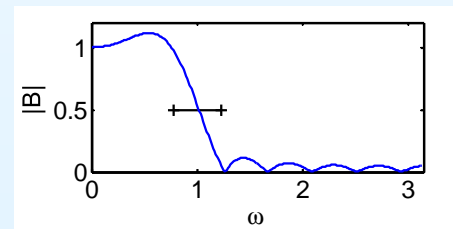
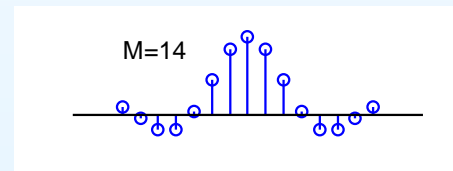
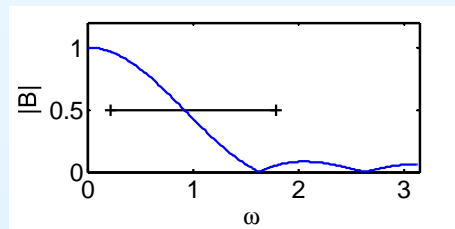
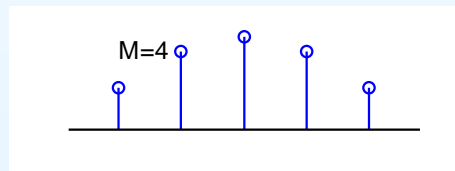
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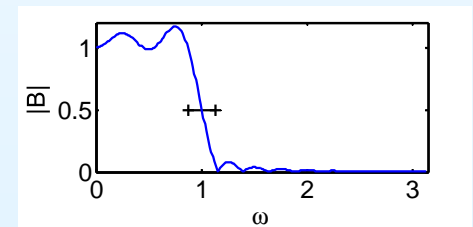
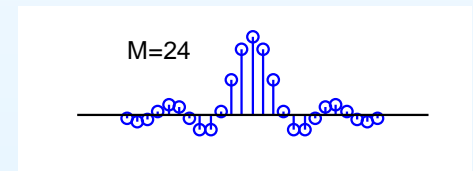
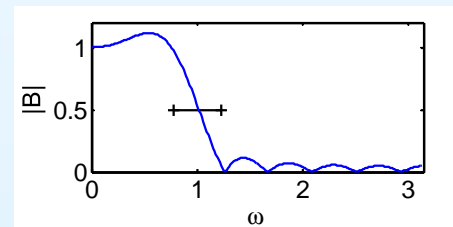
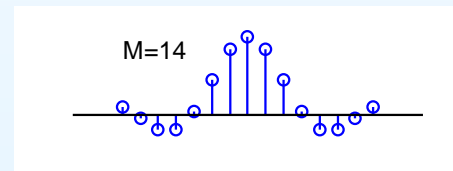
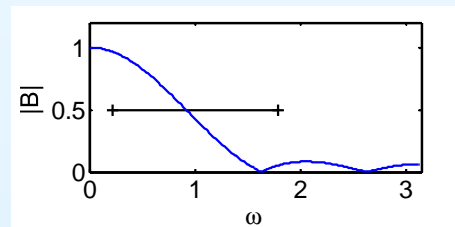
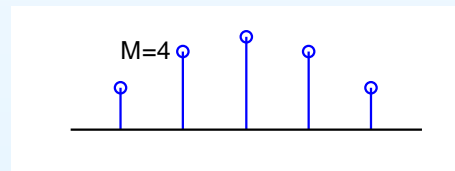
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Rule of thumb: Fastest possible transition $\Delta\omega \geq \frac{2\pi}{M}$ (marked line)

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$B(e^{j\omega})$ is determined by the zeros of $z^M B(z) = \sum_{r=0}^M b[M-r]z^r$

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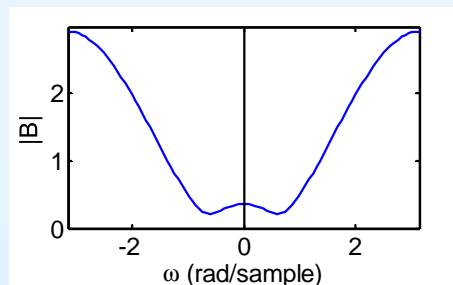
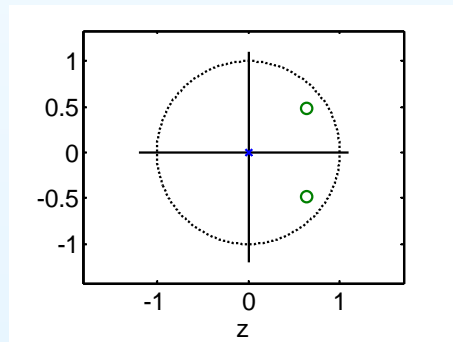
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$B(e^{j\omega})$ is determined by the zeros of $z^M B(z) = \sum_{r=0}^M b[M-r]z^r$

Real $b[n] \Rightarrow$ conjugate zero pairs: $z \Rightarrow z^*$

Real:

[1, -1.28, 0.64]



FIR Symmetries



5: Filters

- Difference Equations
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- **FIR Symmetries** +
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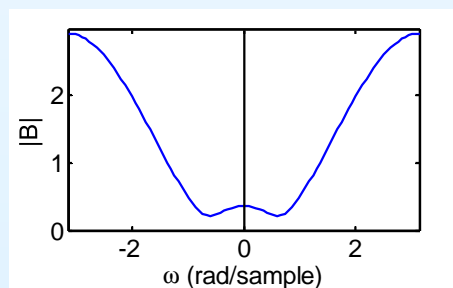
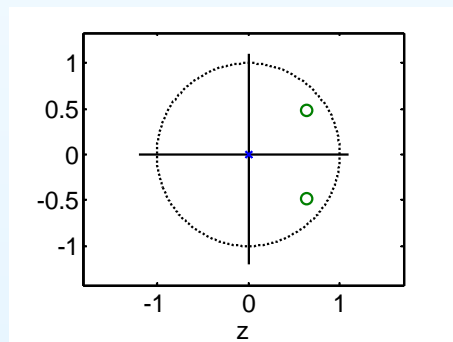
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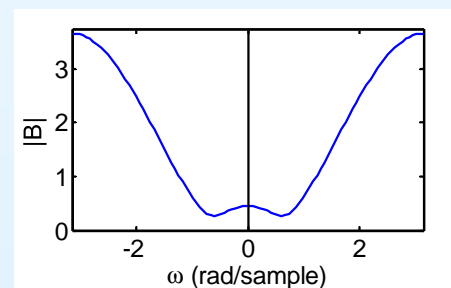
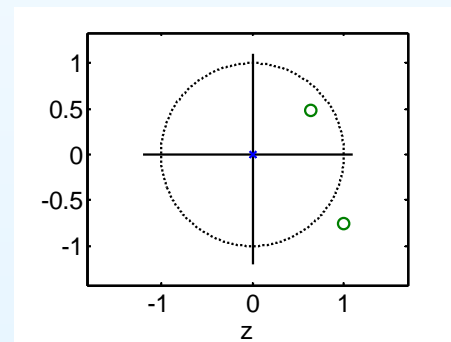
Real:

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Symmetric:

$[1, -1.64 + 0.27j, 1]$



FIR Symmetries



5: Filters

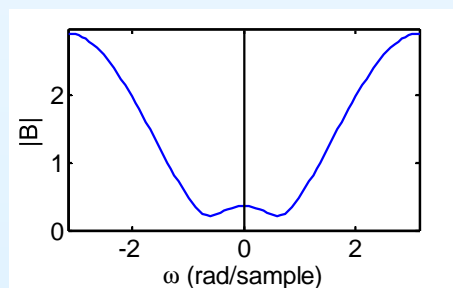
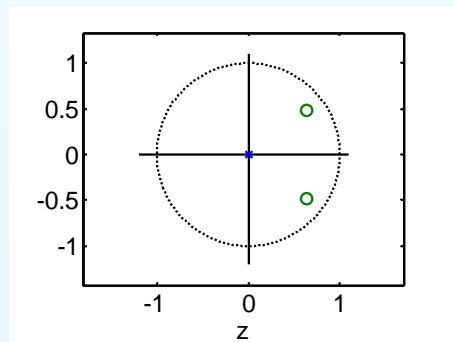
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Real + Symmetric $b[n]$ \Rightarrow conjugate+reciprocal groups of four or else pairs on the real axis

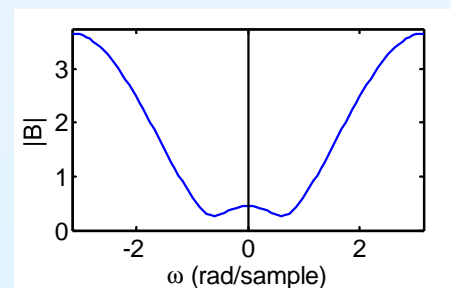
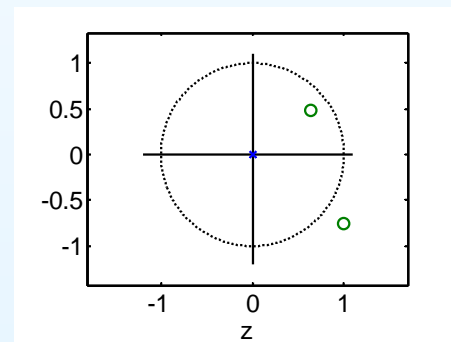
Real:

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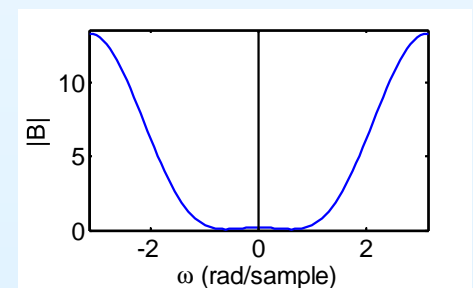
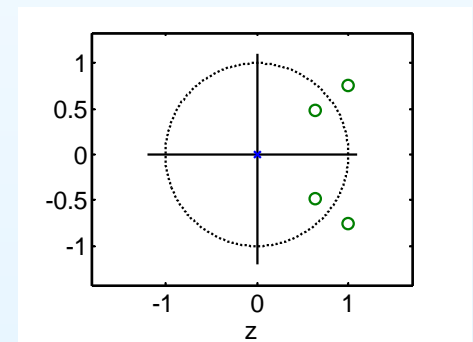
Symmetric:

$[1, -1.64 + 0.27j, 1]$



Real + Symmetric:

$[1, -3.28, 4.7625, -3.28, 1]$



IIR Frequency Response

5: Filters

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$$\text{Factorize } H(z) = \frac{B(z)}{A(z)} = \frac{b[0] \prod_{i=1}^M (1 - q_i z^{-1})}{\prod_{i=1}^N (1 - p_i z^{-1})}$$

IIR Frequency Response

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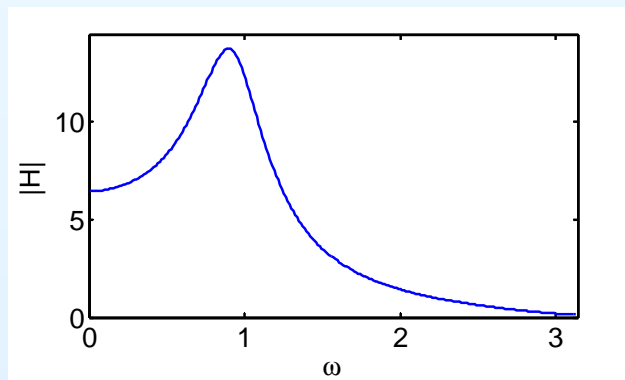
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Example:

$$H(z) = \frac{2 + 2.4z^{-1}}{1 - 0.96z^{-1} + 0.64z^{-2}}$$



IIR Frequency Response

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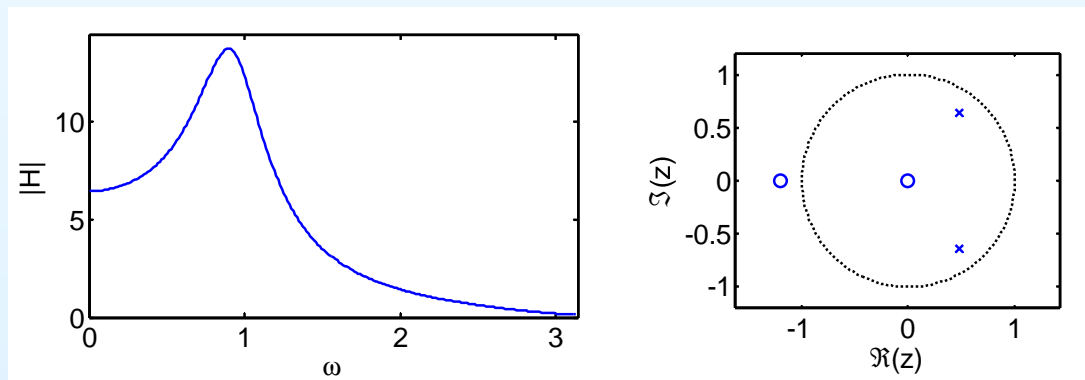
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Example:

$$H(z) = \frac{2 + 2.4z^{-1}}{1 - 0.96z^{-1} + 0.64z^{-2}} = \frac{2(1 + 1.2z^{-1})}{(1 - (0.48 - 0.64j)z^{-1})(1 - (0.48 + 0.64j)z^{-1})}$$



IIR Frequency Response

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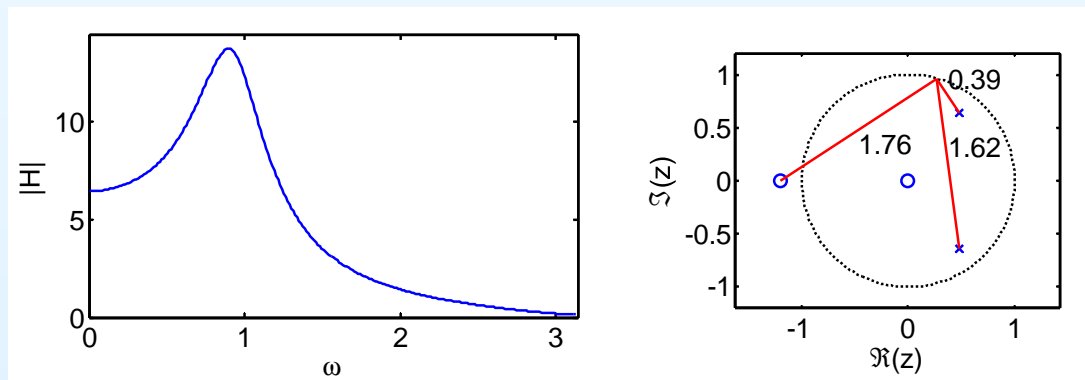
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$$\text{At } \omega = 1.3: |H(e^{j\omega})| = \frac{2 \times 1.76}{1.62 \times 0.39}$$



IIR Frequency Response

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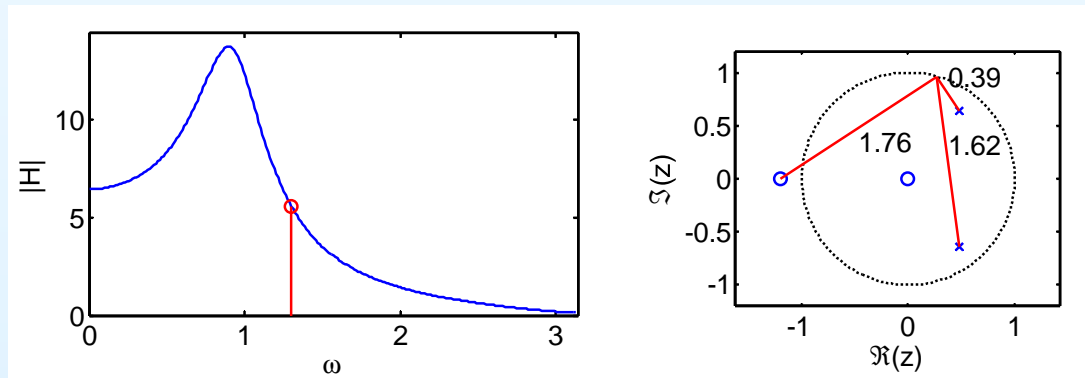
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At $\omega = 1.3$: $|H(e^{j\omega})| = \frac{2 \times 1.76}{1.62 \times 0.39} = 5.6$



IIR Frequency Response

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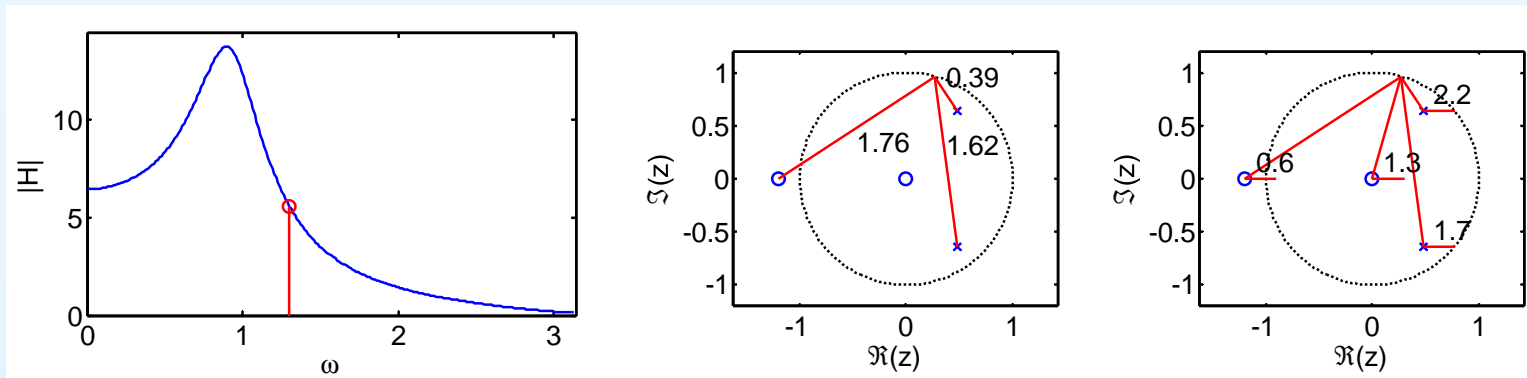
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At $\omega = 1.3$: $|H(e^{j\omega})| = \frac{2 \times 1.76}{1.62 \times 0.39} = 5.6$

$$\angle H(e^{j\omega}) = (0.6 + 1.3) - (1.7 + 2.2) = -1.97$$



Negating z

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Given a filter $H(z)$ we can form a new one $H_R(z) = H(-z)$

Negating z

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Negate all odd powers of z , i.e. negate alternate $a[n]$ and $b[n]$

Negating z

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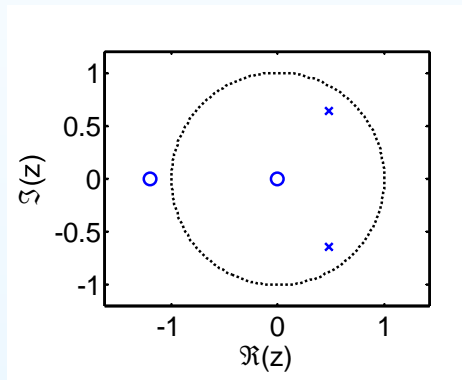
Negating z

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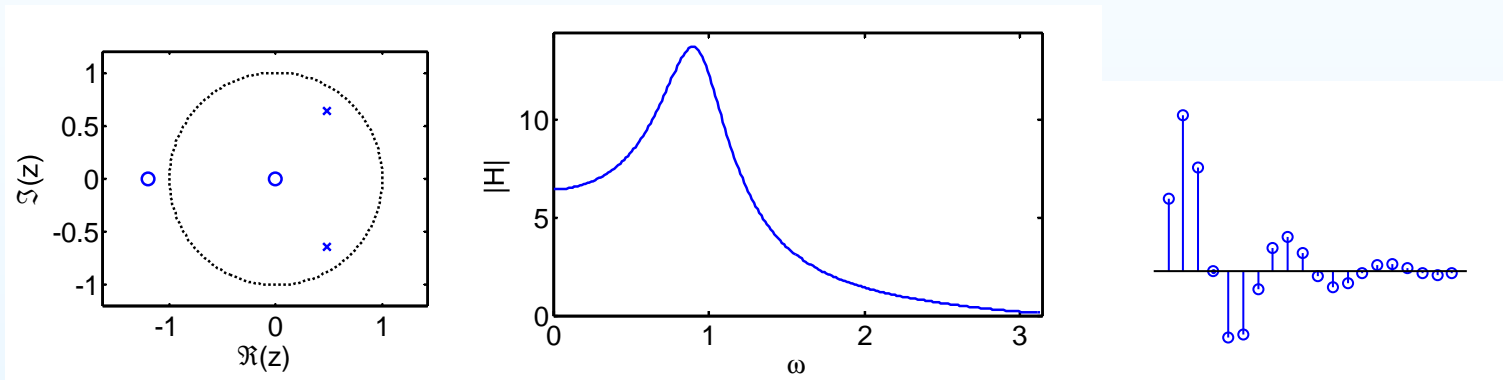
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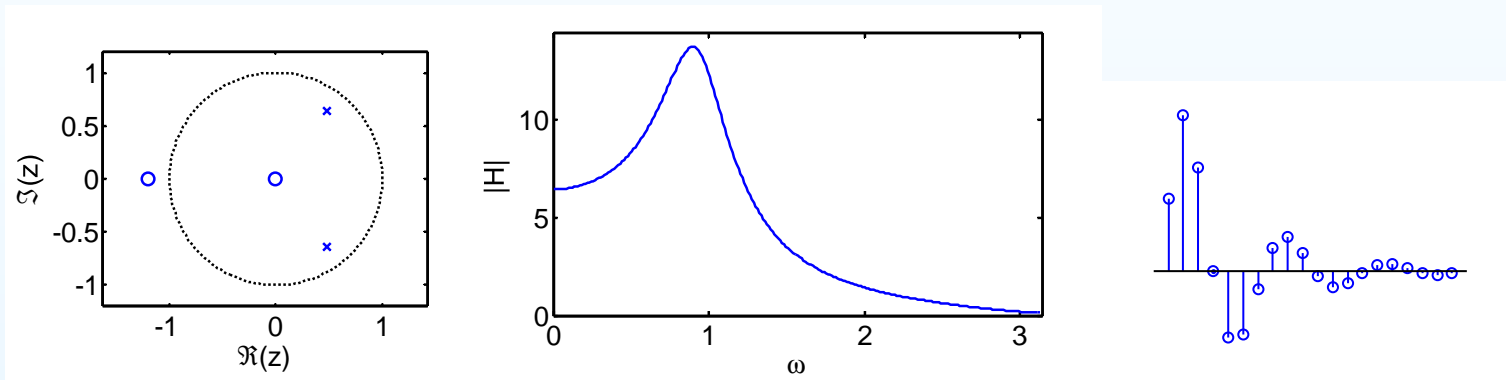
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Negate z:
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Negate odd coefficients

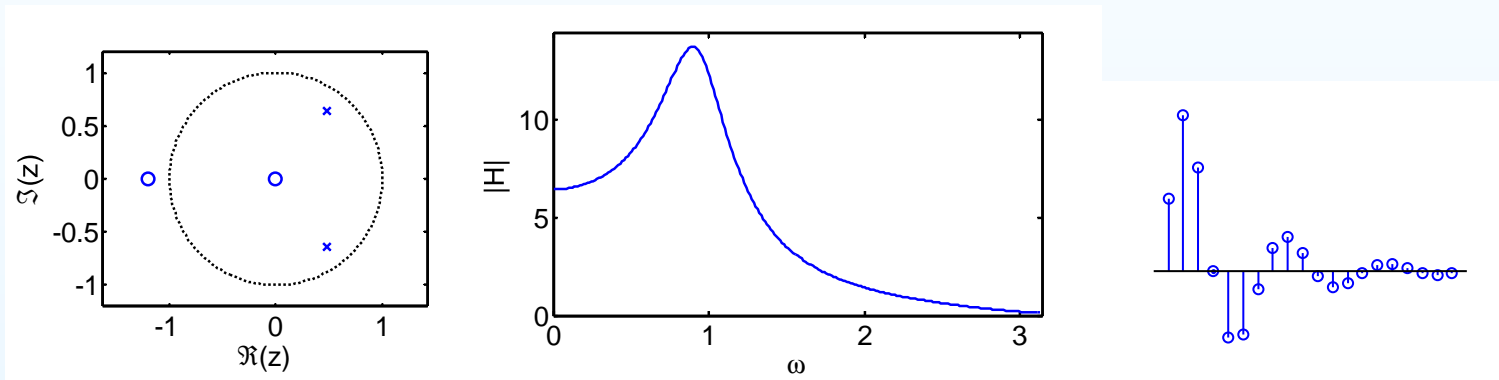
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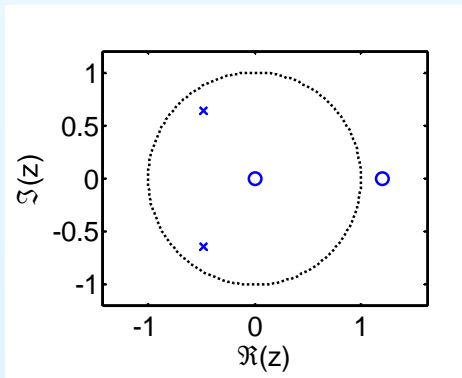
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Negate odd coefficients



Pole and zero positions are **negated**

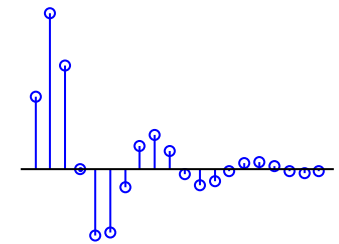
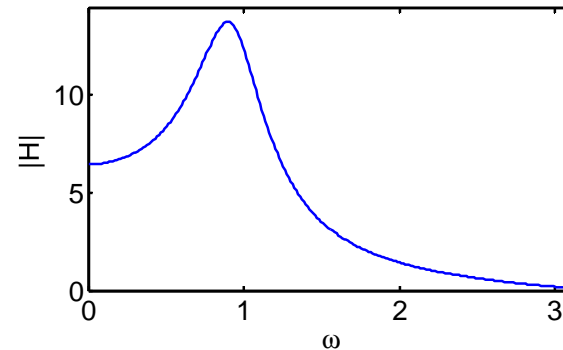
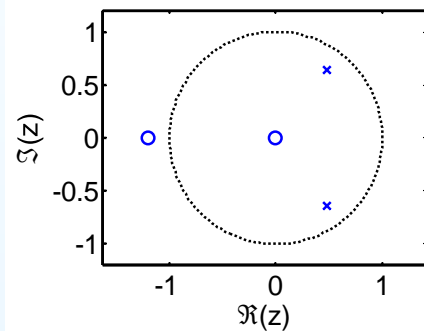
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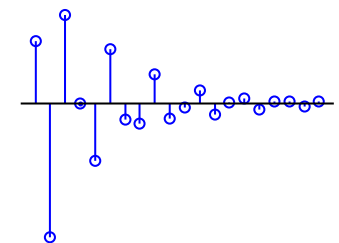
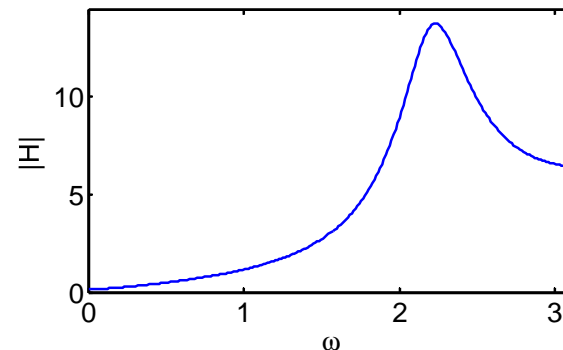
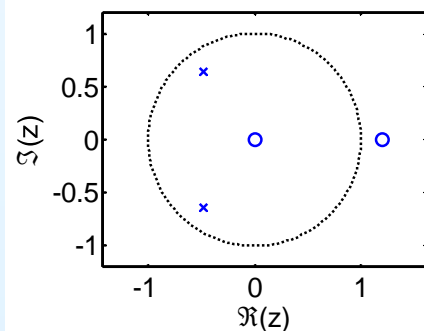
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Negate z :
$$H_R(z) = \frac{2-2.4z^{-1}}{1+0.96z^{-1}+0.64z^{-2}}$$

Negate odd coefficients



Pole and zero positions are **negated**, response is **flipped** and **conjugated**.

Cubing z



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Given a filter $H(z)$ we can form a new one $H_C(z) = H(z^3)$

Cubing z

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Given a filter $H(z)$ we can form a new one $H_C(z) = H(z^3)$
Insert two zeros between each $a[n]$ and $b[n]$ term

Cubing z

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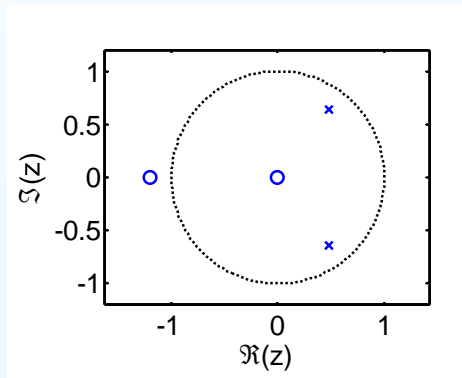
Cubing z

5: Filters

- Difference Equations
- FIR Filters
- FIR Symmetries +
- IIR Frequency Response
- Negating z +
- **Cubing z** +
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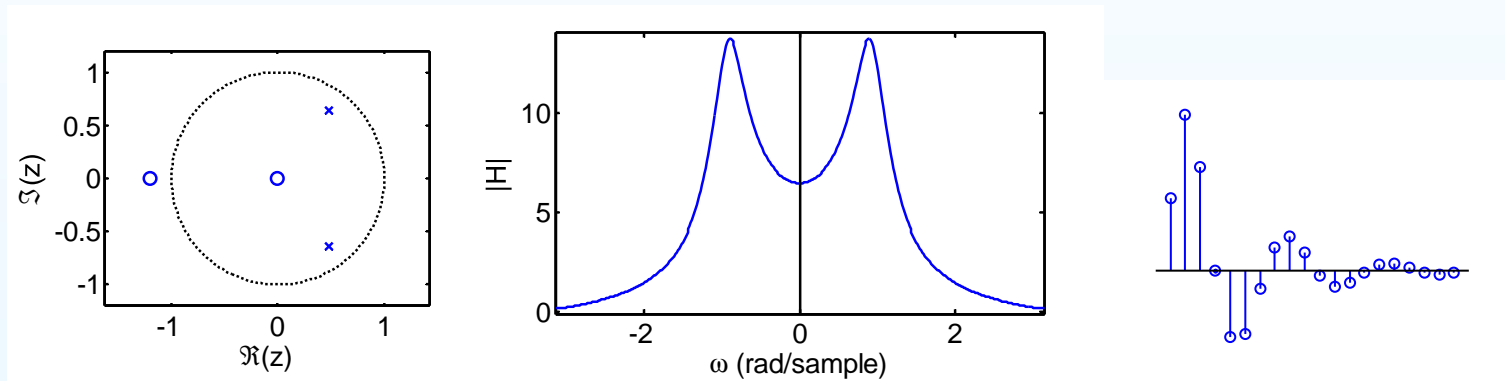
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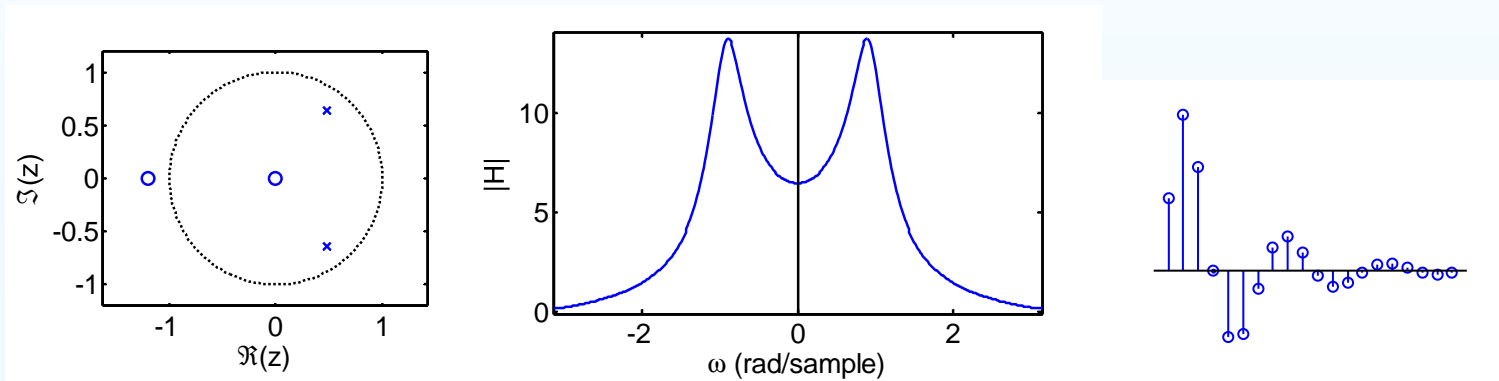
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Insert 2 zeros between coeffs

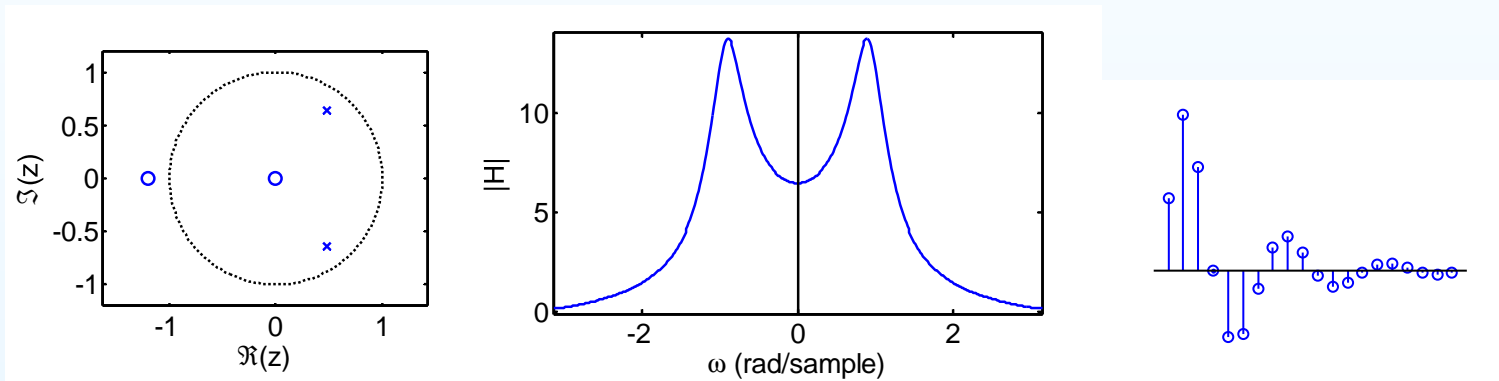
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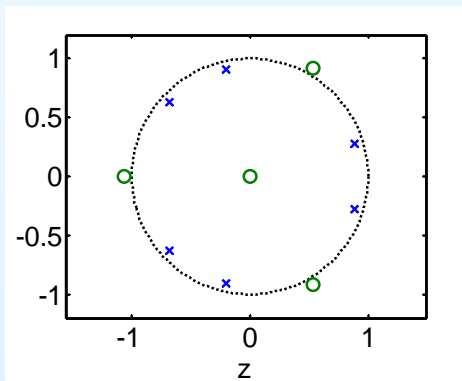
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Pole and zero positions are **replicated**

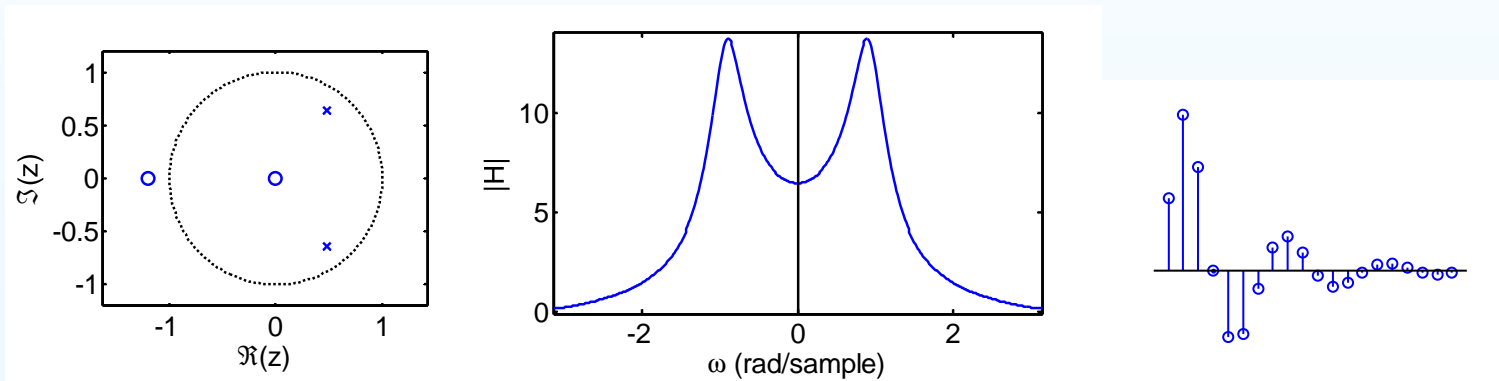
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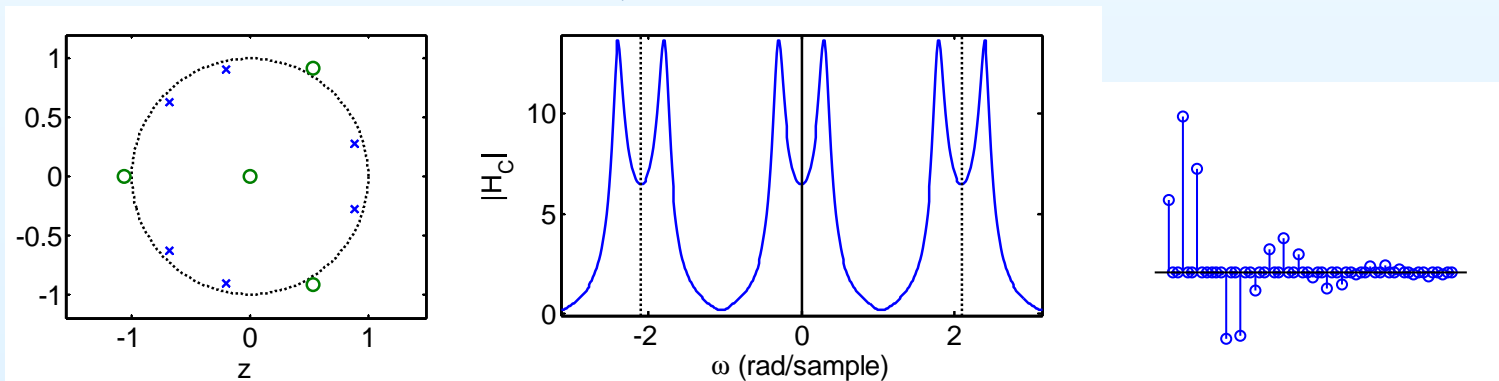
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Pole and zero positions are **replicated**, magnitude response **replicated**.

Scaling z



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Scaling z

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Scaling z

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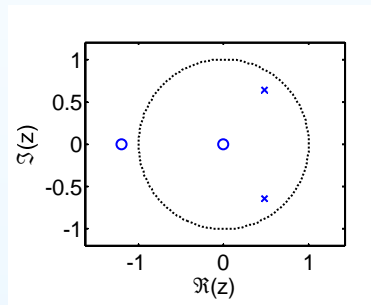
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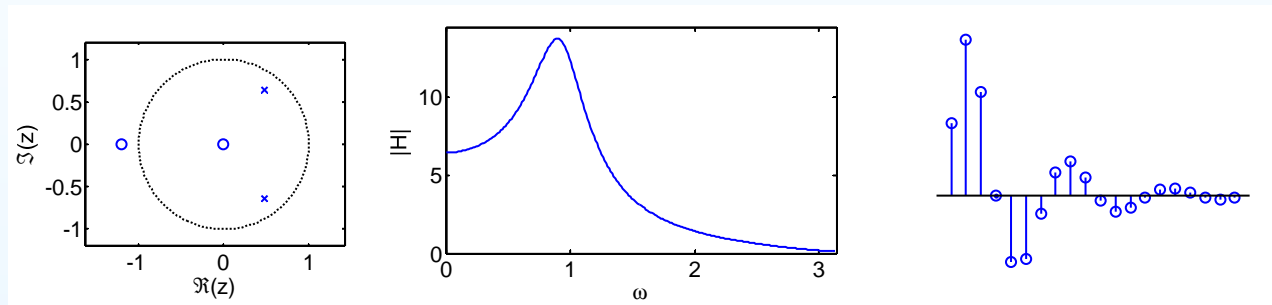
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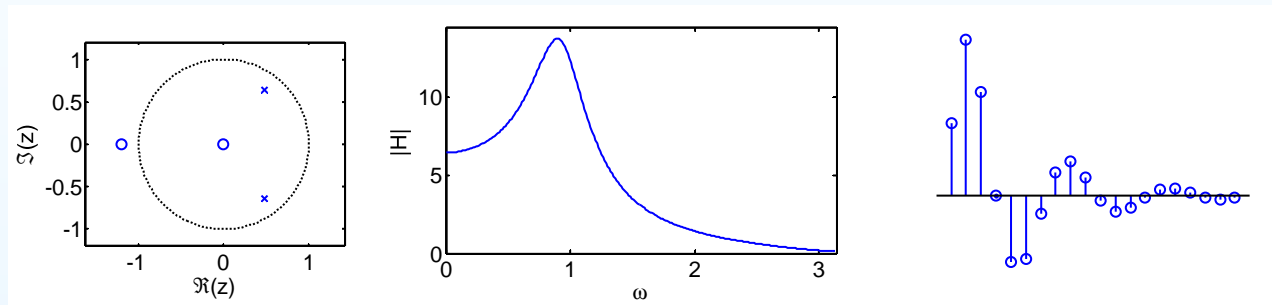
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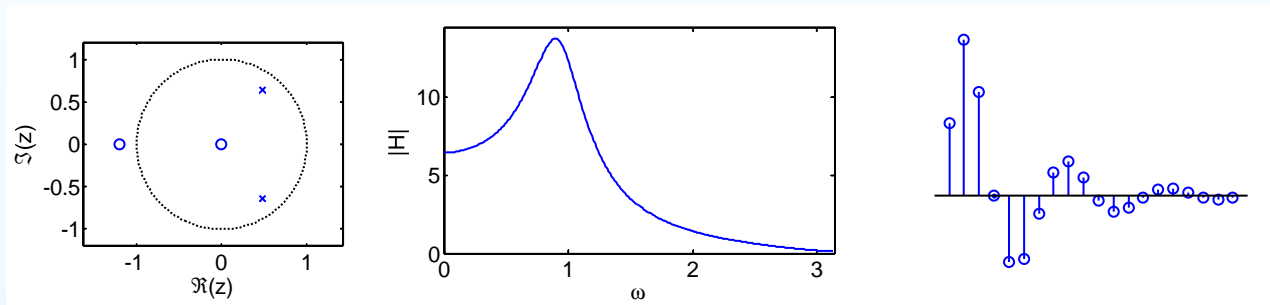
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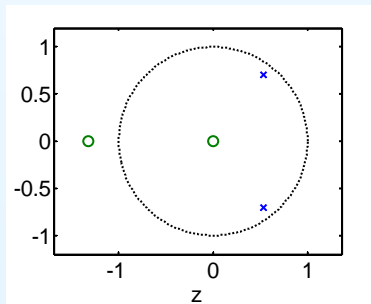
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Pole and zero positions are **multiplied by α**

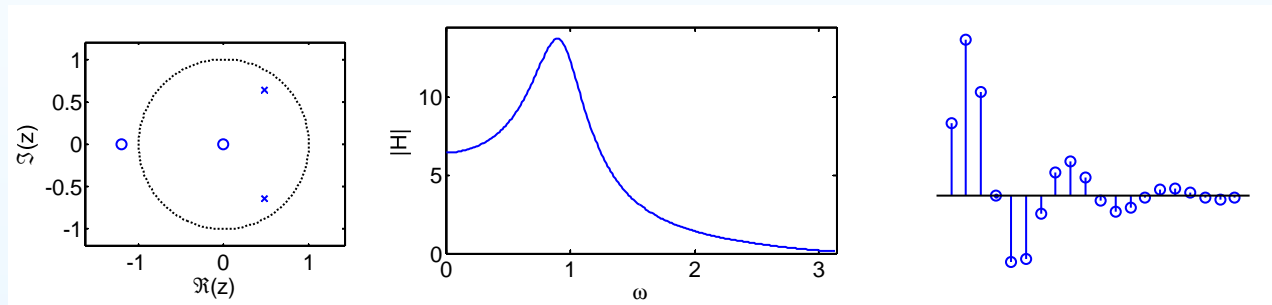
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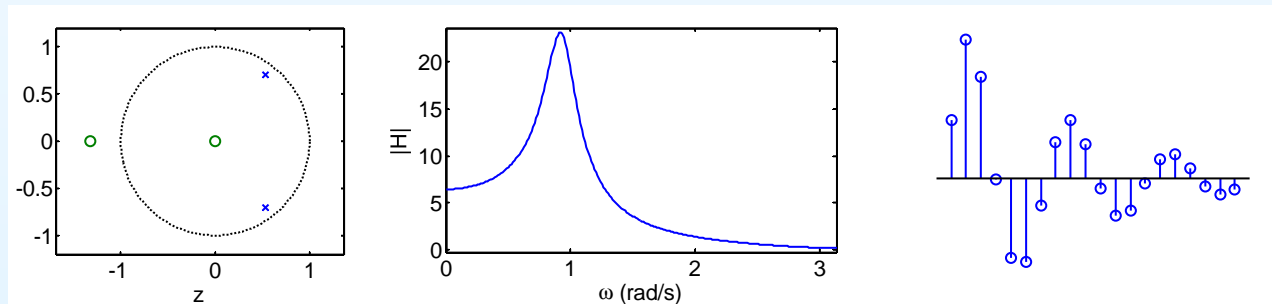
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Pole and zero positions are **multiplied by α** , $\alpha > 1 \Rightarrow$ peaks **sharpened**.

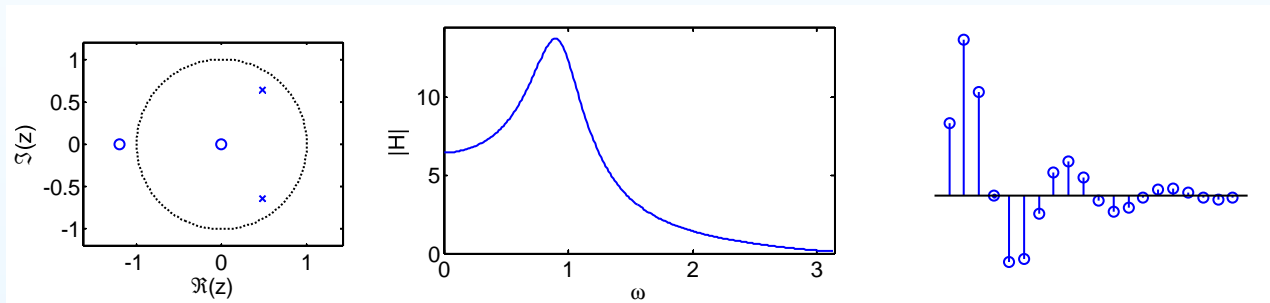
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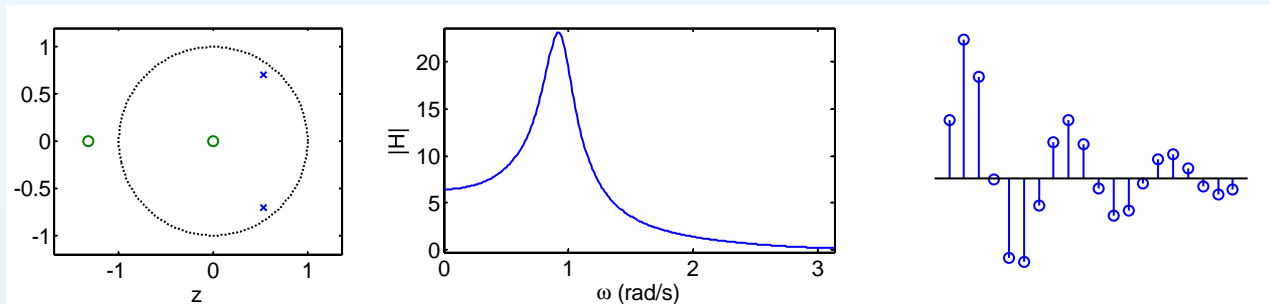
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Pole at $z = p$ gives peak bandwidth $\approx 2 |\log |p|| \approx 2 (1 - |p|)$

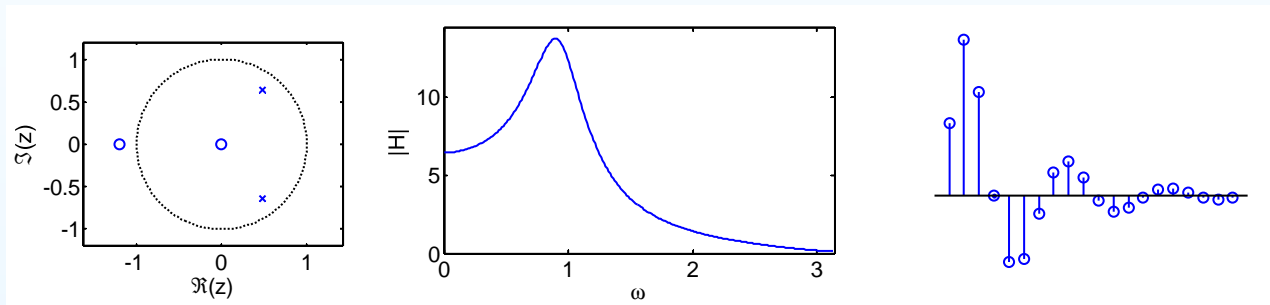
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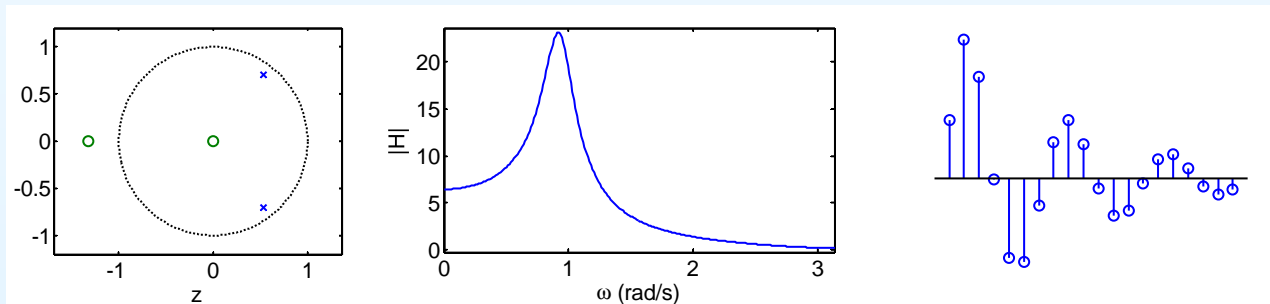
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Pole at $z = p$ gives peak bandwidth $\approx 2 |\log |p|| \approx 2 (1 - |p|)$

For pole near unit circle, **decrease bandwidth** by $\approx 2 \log \alpha$

Low-pass filter



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1st order low pass filter: extremely common

$$y[n] = (1 - p)x[n] + py[n - 1]$$

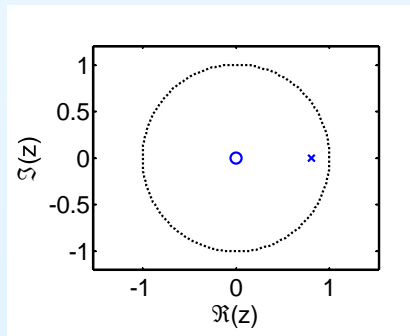
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Low-pass filter

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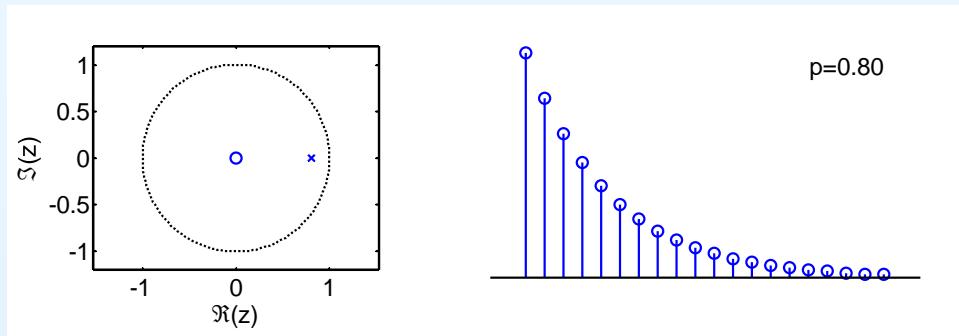
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Impulse response:

$$h[n] = (1 - p)p^n$$



Low-pass filter

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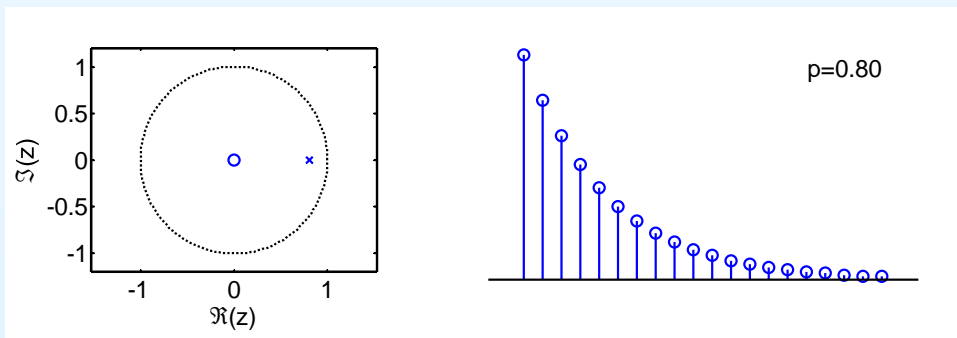
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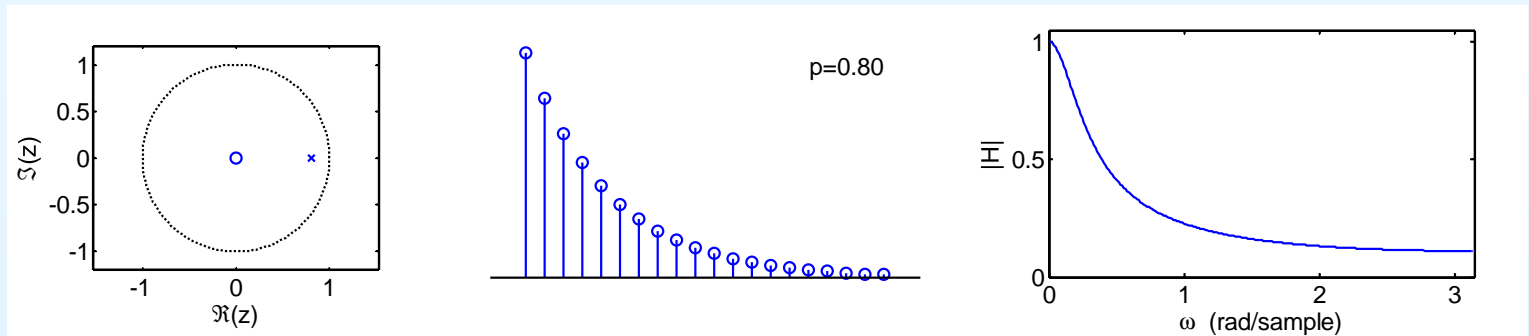
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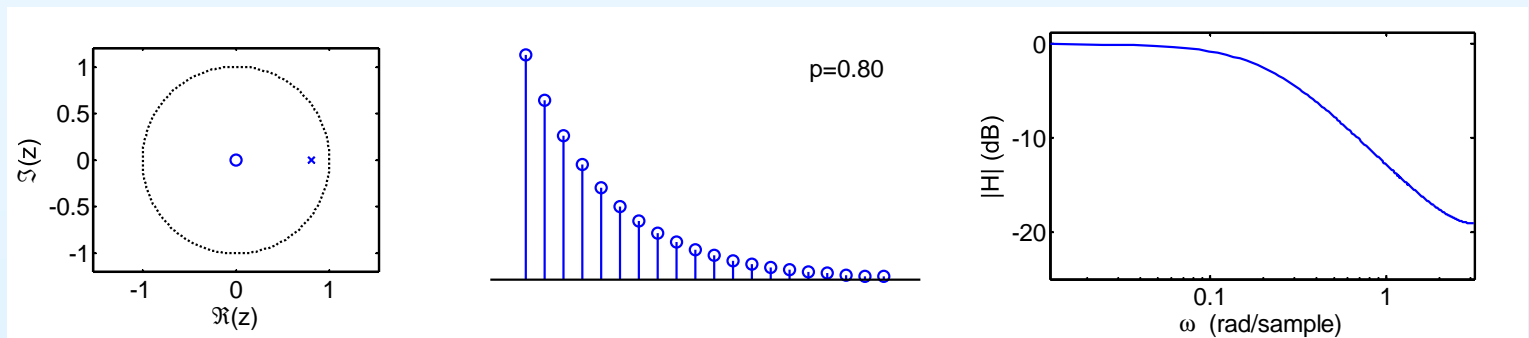
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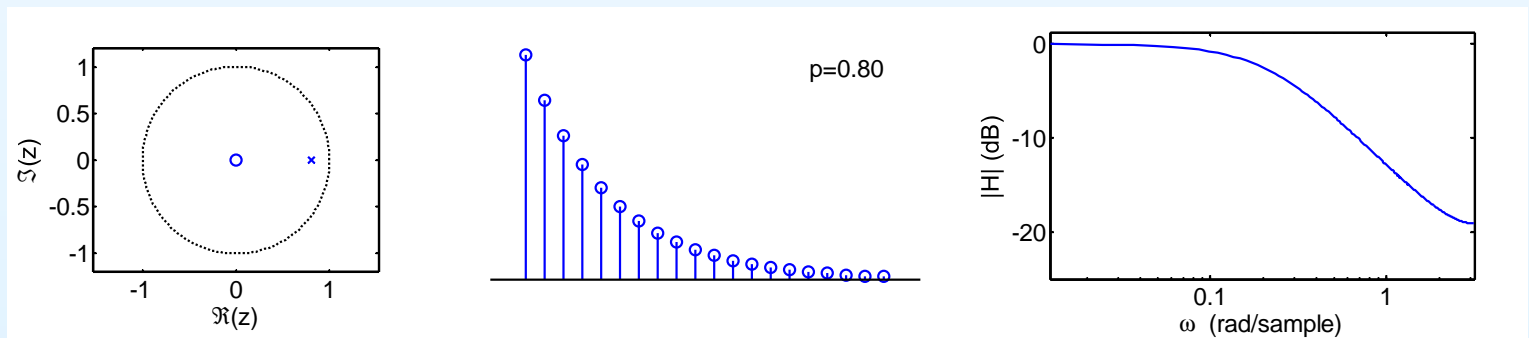
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Low-pass filter with DC gain of unity.



Low-pass filter

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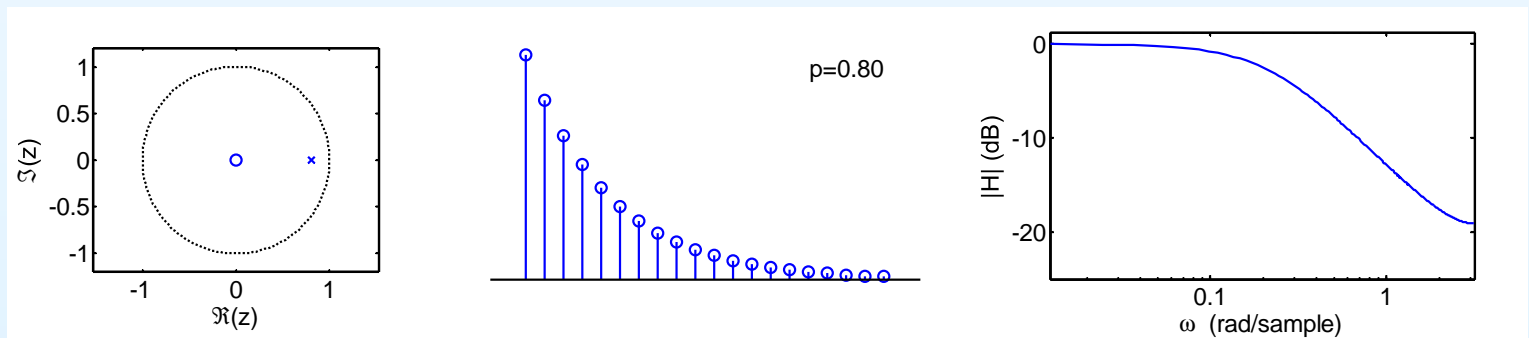
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3 dB frequency is $\omega_{3dB} = \cos^{-1} \left(1 - \frac{(1-p)^2}{2p} \right)$



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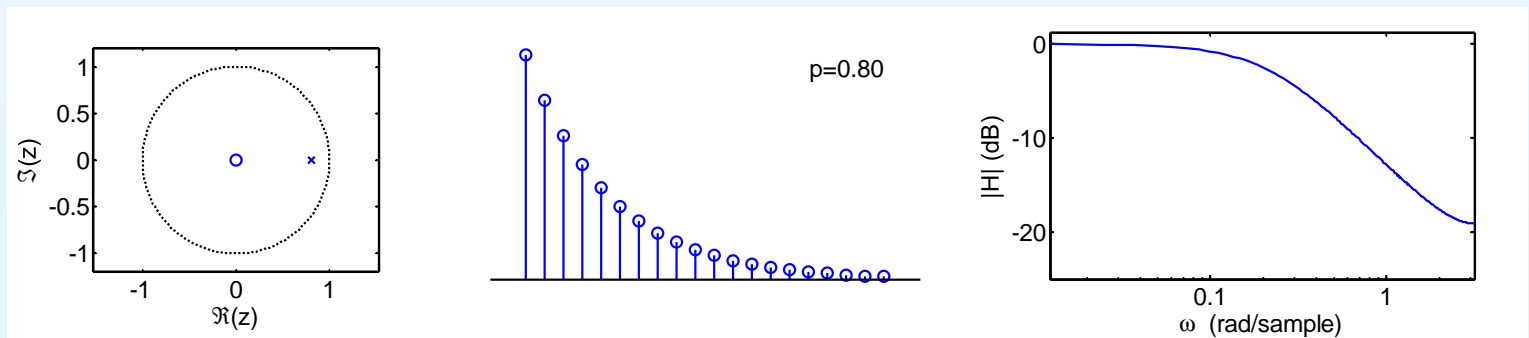
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3 dB frequency is $\omega_{3dB} = \cos^{-1} \left(1 - \frac{(1-p)^2}{2p} \right) \approx 2 \frac{1-p}{1+p}$



Low-pass filter

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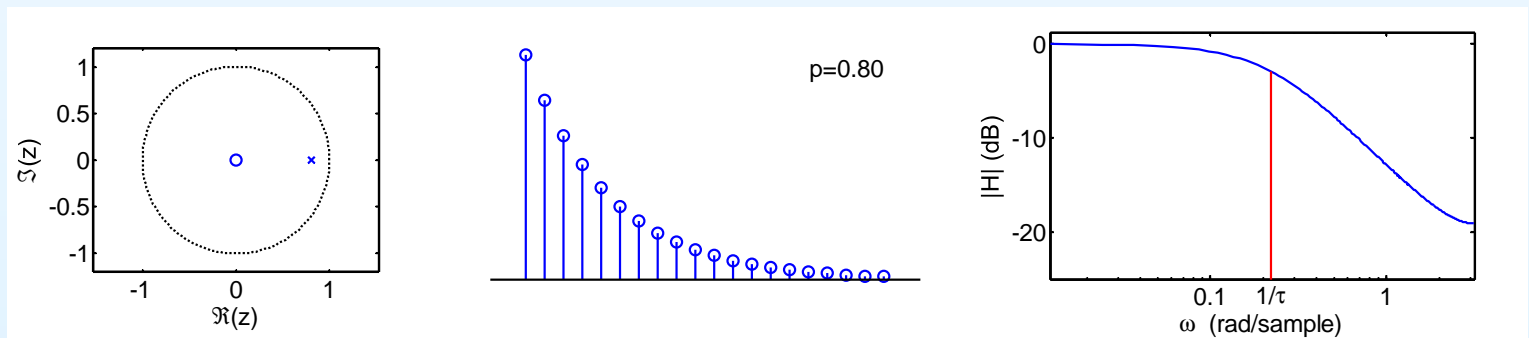
$$h[n] = (1 - p)p^n = (1 - p)e^{-\frac{n}{\tau}}$$

where $\tau = \frac{1}{-\ln p}$ is the time constant in samples.

Magnitude response: $|H(e^{j\omega})| = \frac{1-p}{\sqrt{1-2p \cos \omega + p^2}}$

Low-pass filter with DC gain of unity.

3 dB frequency is $\omega_{3dB} = \cos^{-1} \left(1 - \frac{(1-p)^2}{2p} \right) \approx 2 \frac{1-p}{1+p} \approx \frac{1}{\tau}$



Low-pass filter



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1st order low pass filter: extremely common

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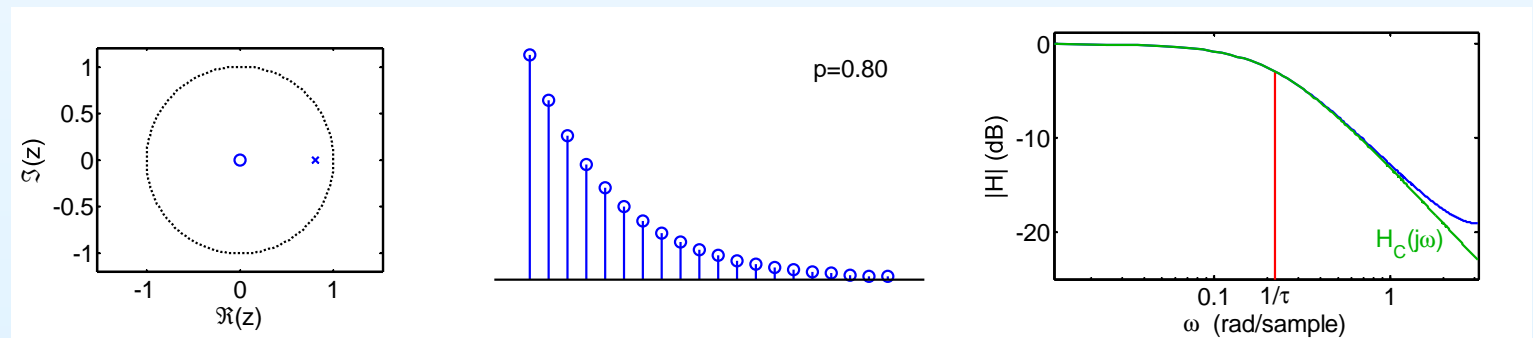
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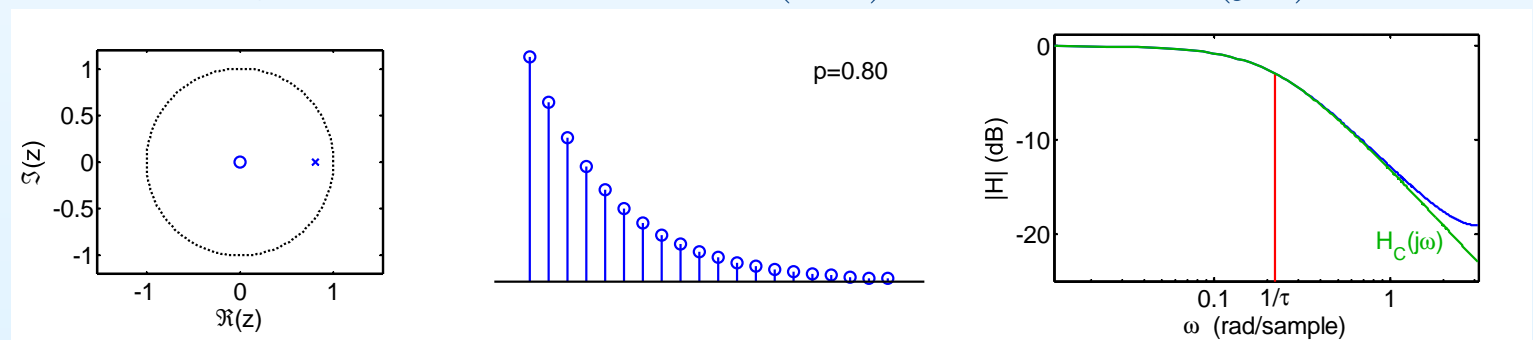
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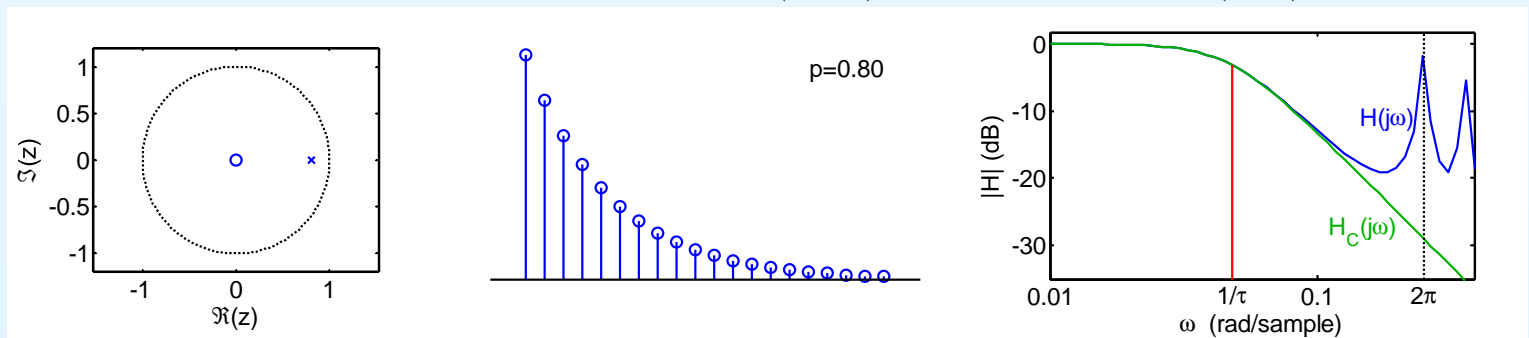
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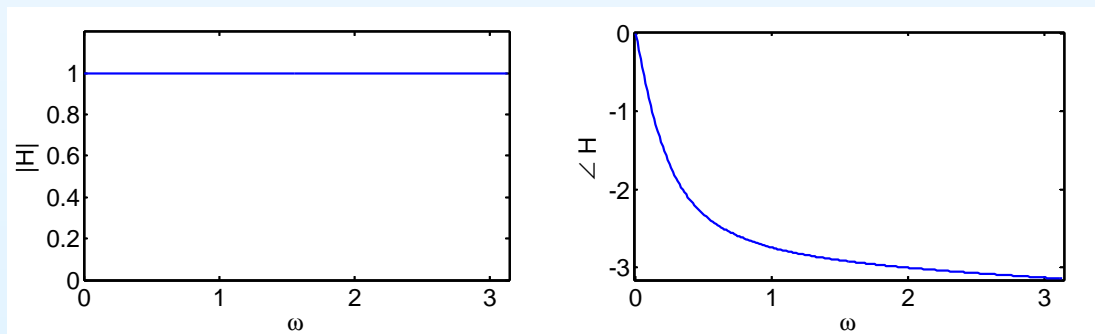
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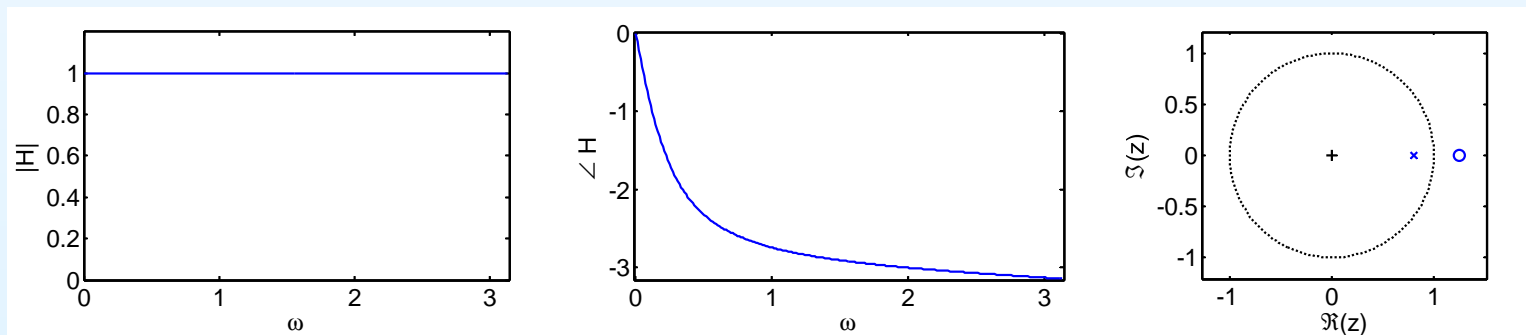
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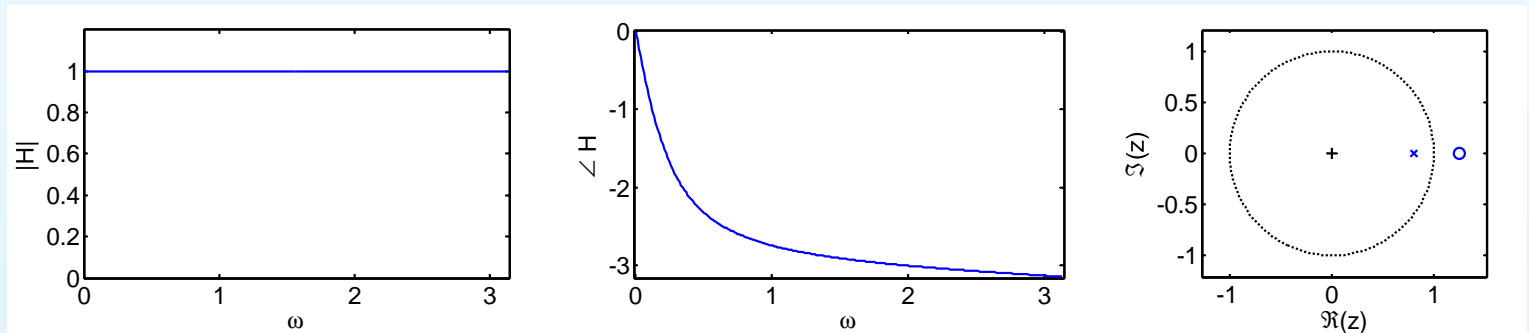
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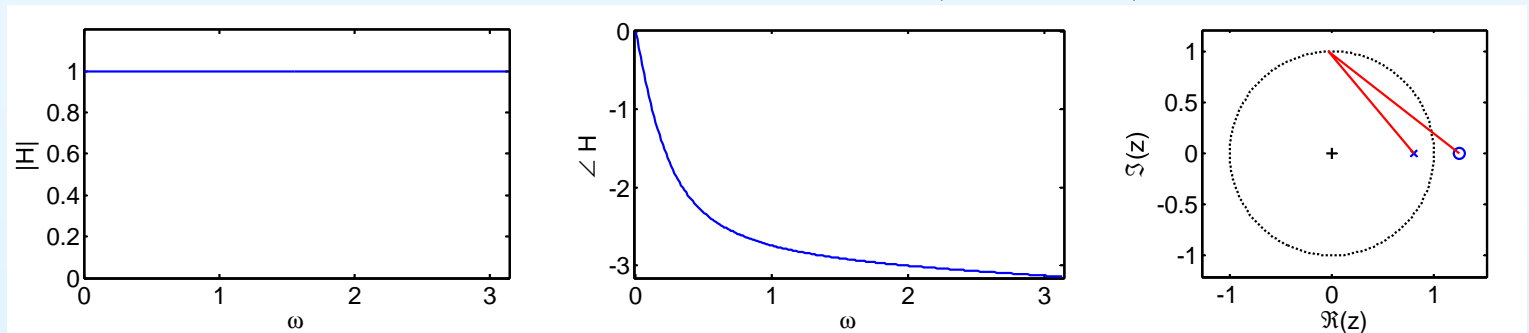
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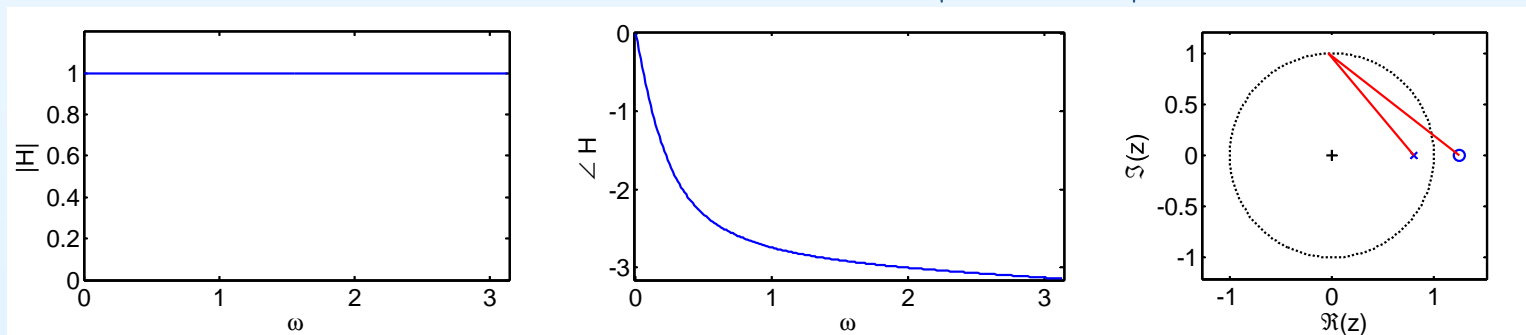
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In an allpass filter, the **zeros are the poles reflected in the unit circle**.

Group Delay



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Example: $H(z) = \frac{1}{1-pz^{-1}}$

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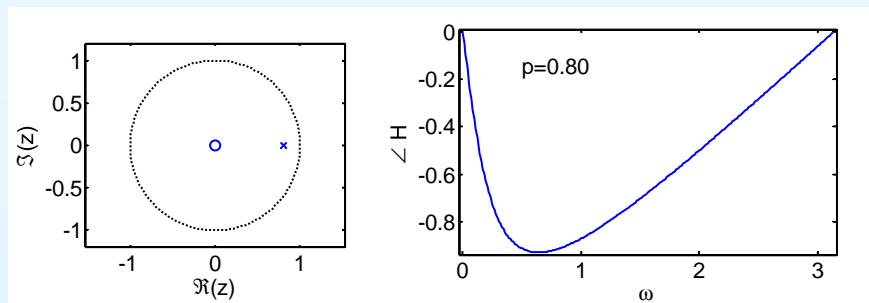
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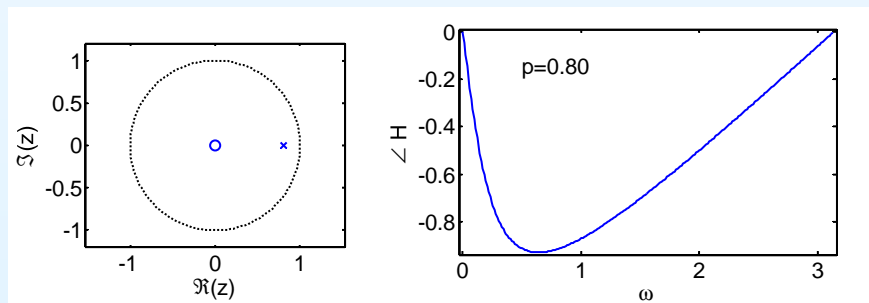
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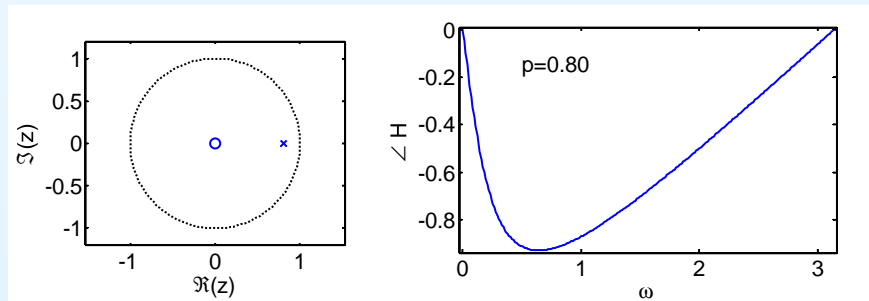
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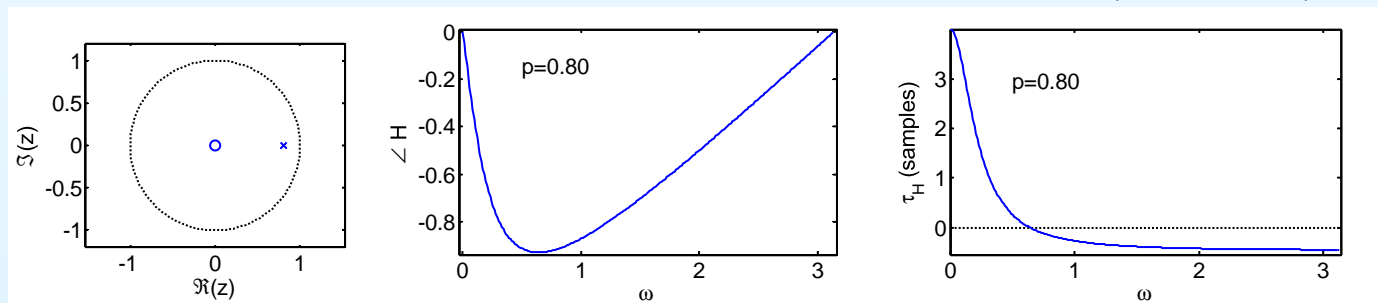
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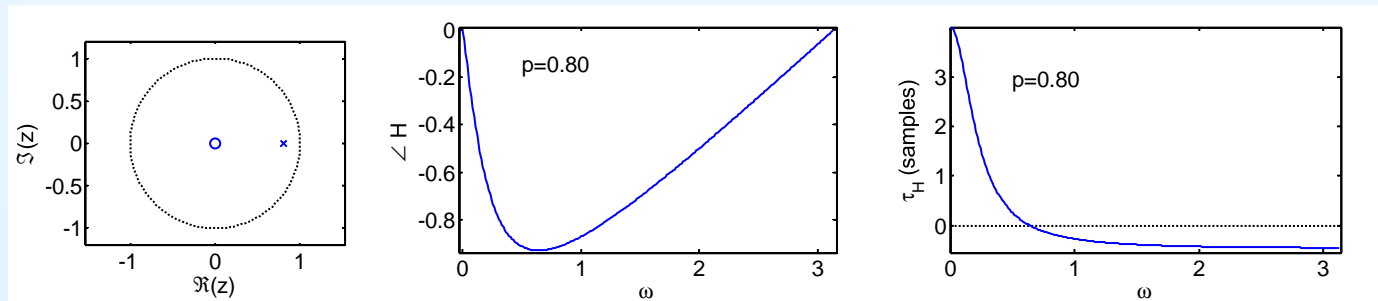
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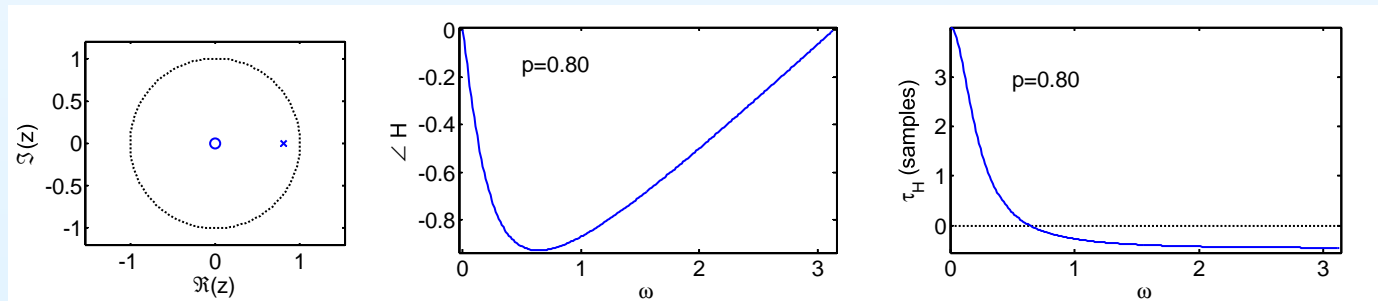
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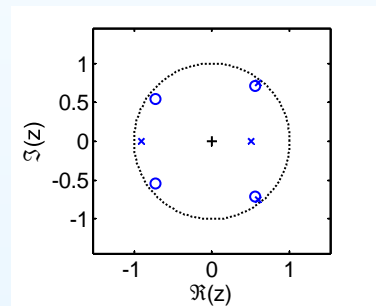
Zeros on the unit circle count $-1/2$

Minimum Phase

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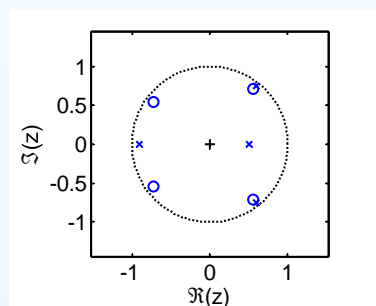
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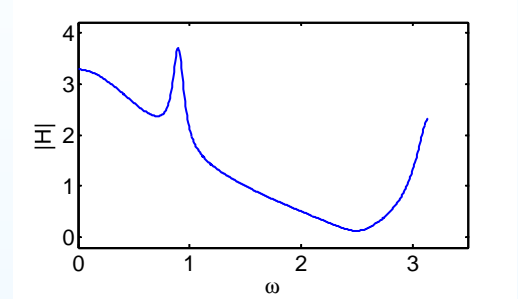
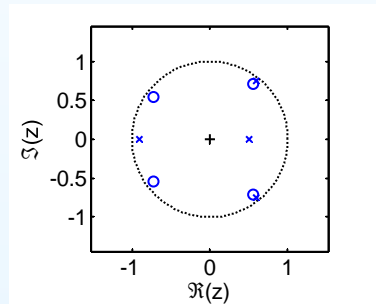
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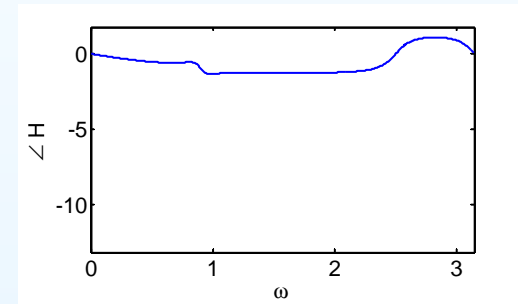
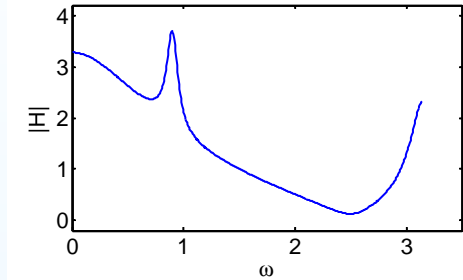
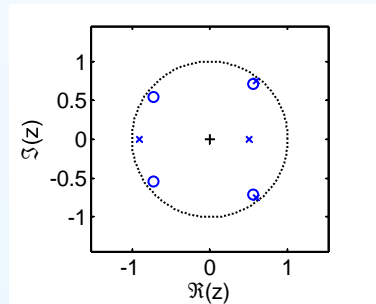
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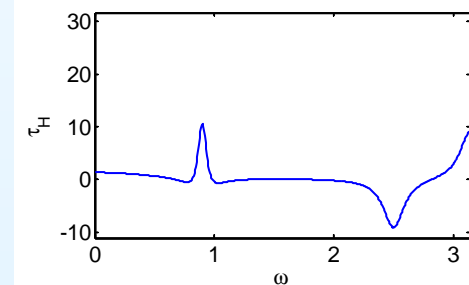
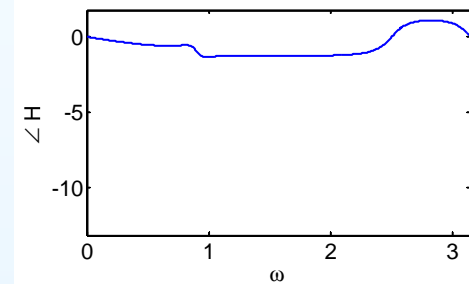
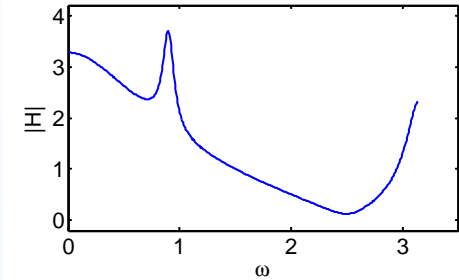
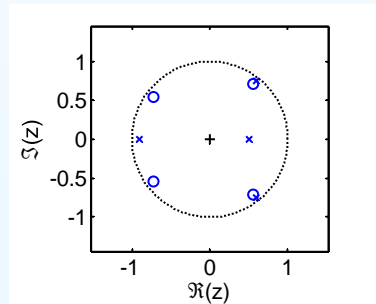
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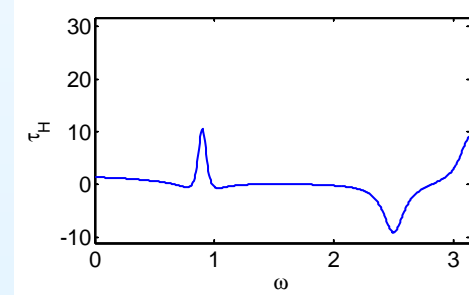
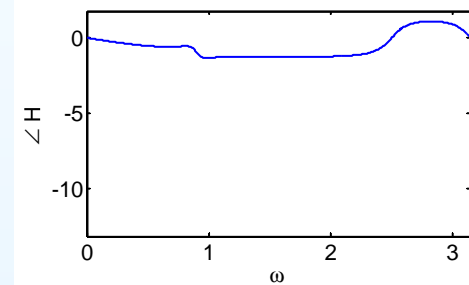
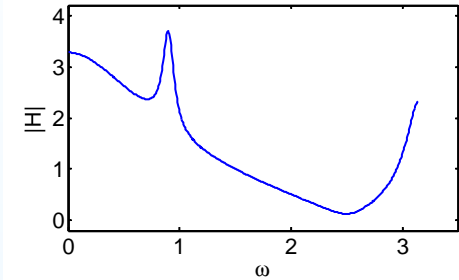
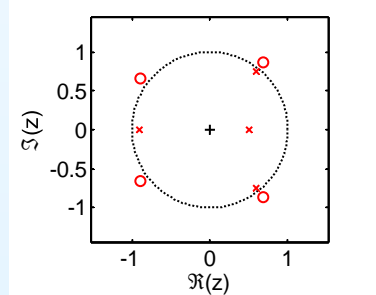
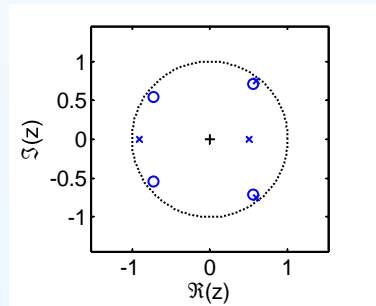


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Minimum Phase

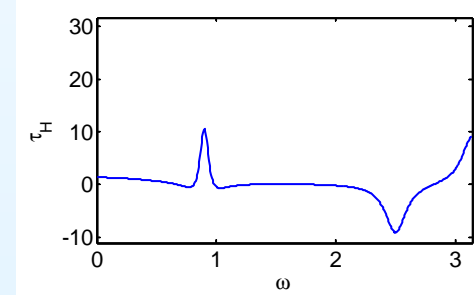
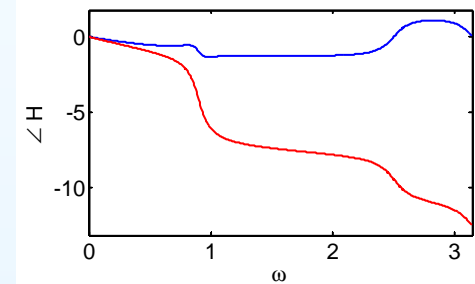
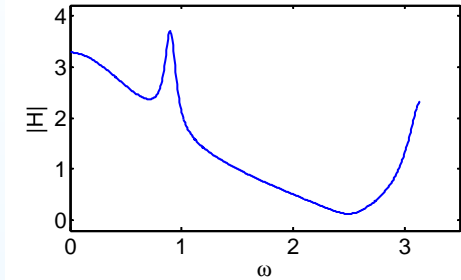
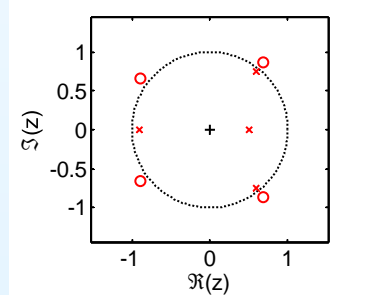
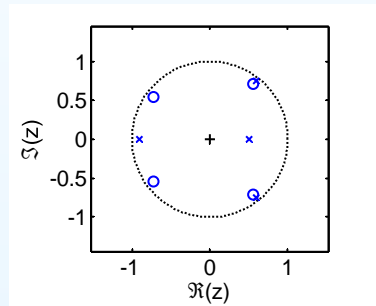


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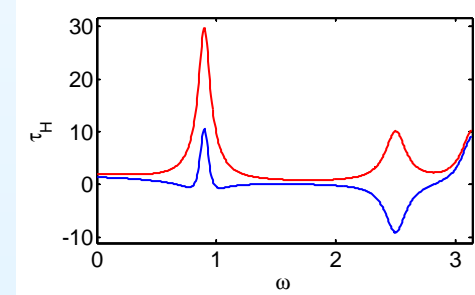
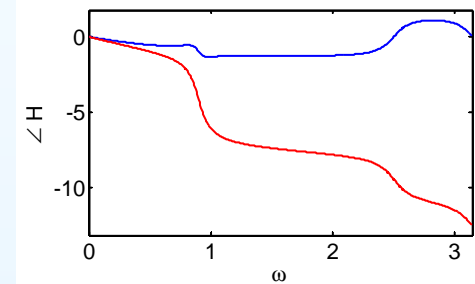
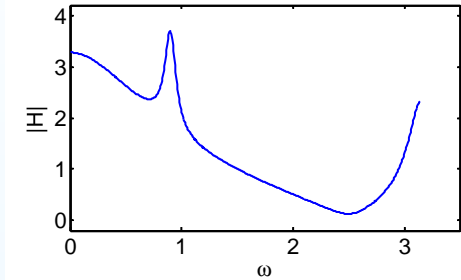
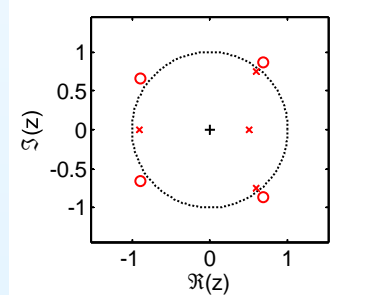
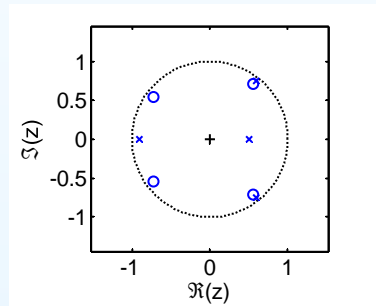


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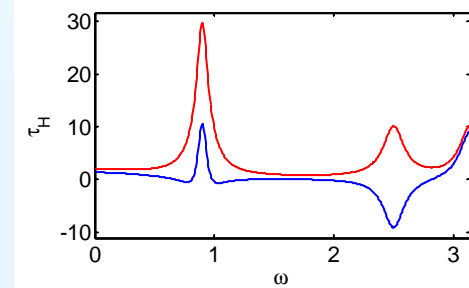
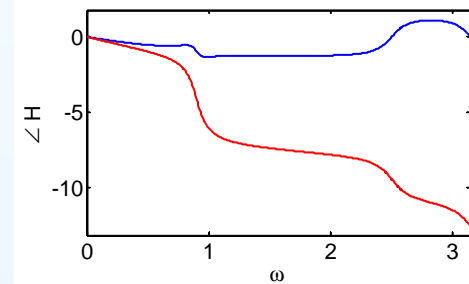
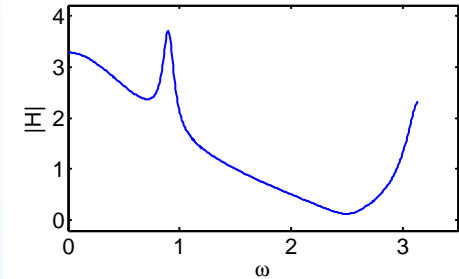
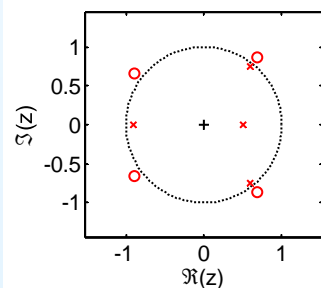
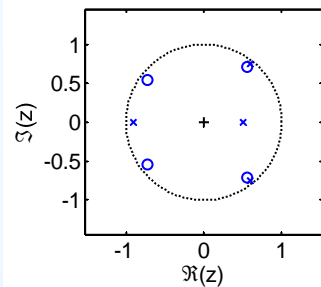
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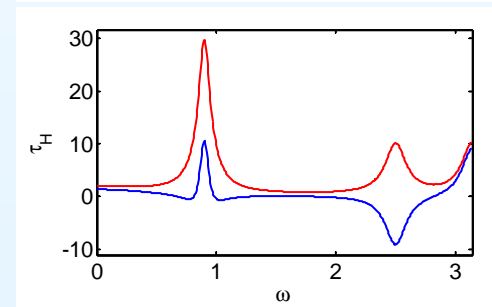
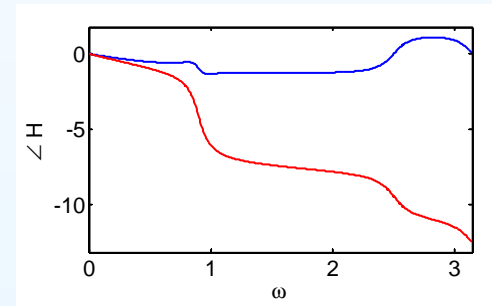
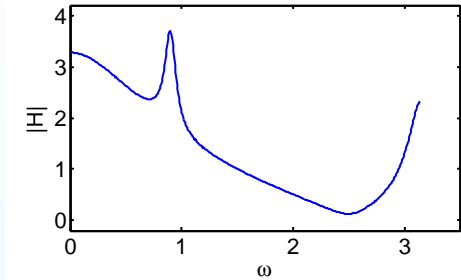
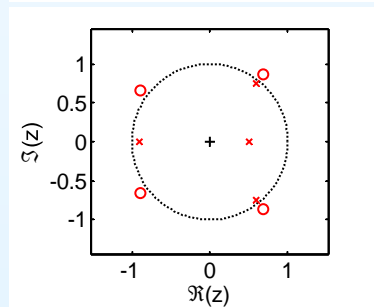
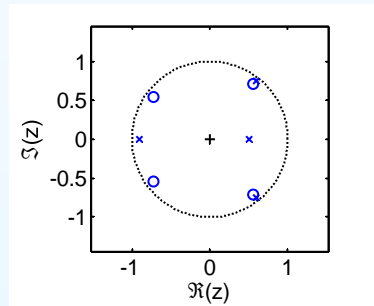
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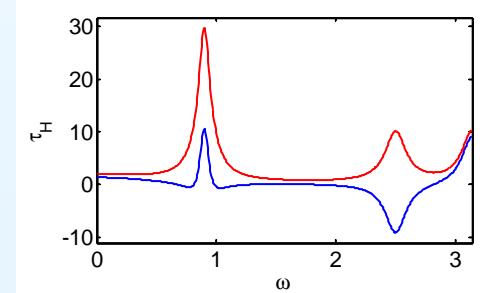
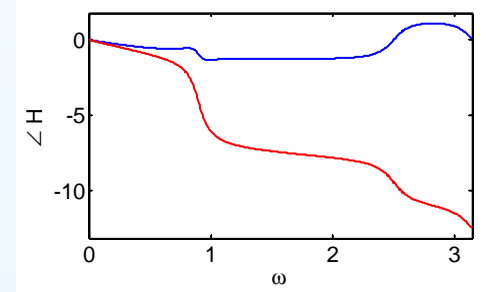
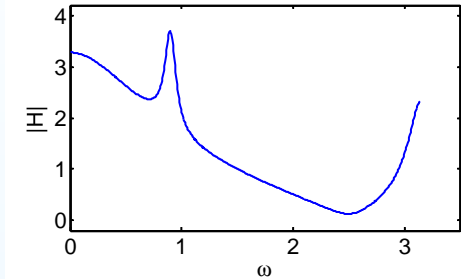
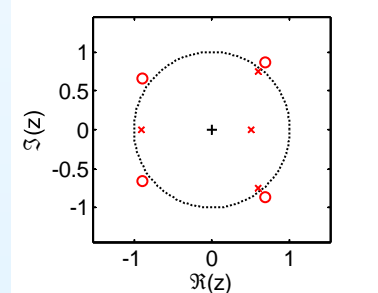
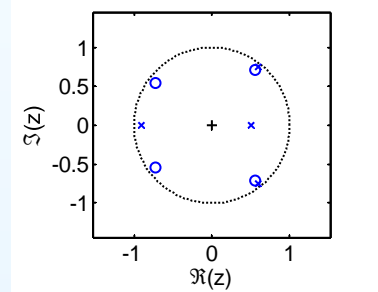
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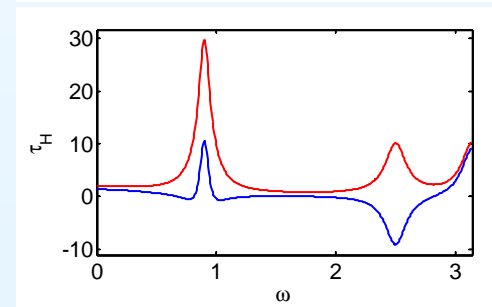
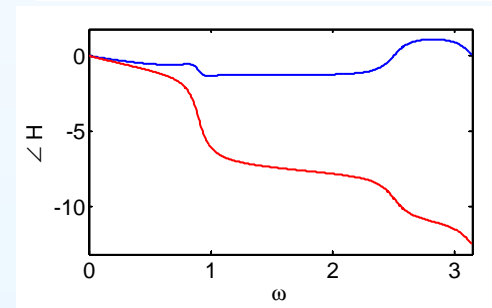
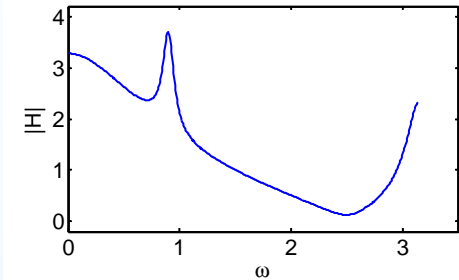
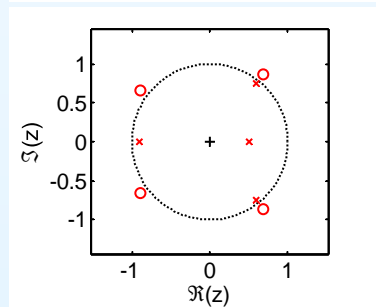
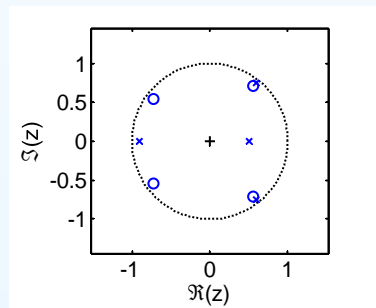
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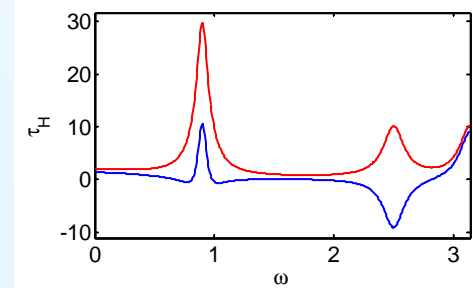
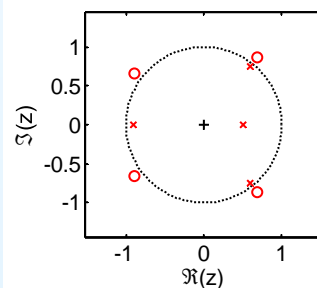
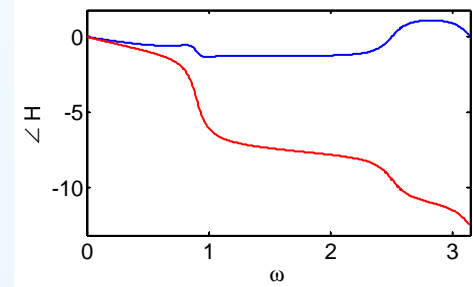
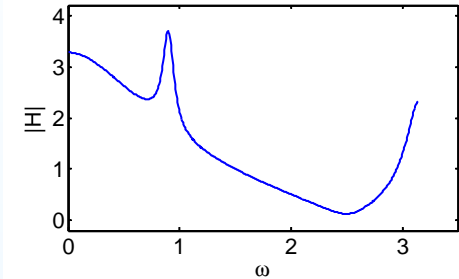
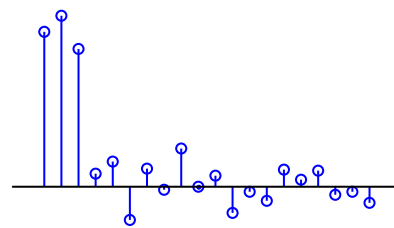
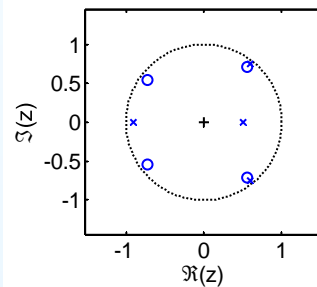
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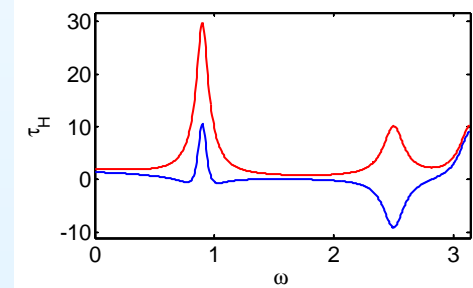
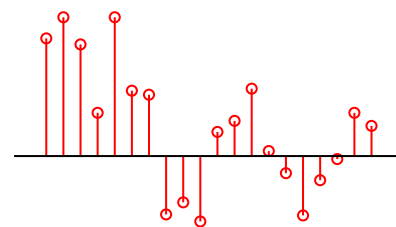
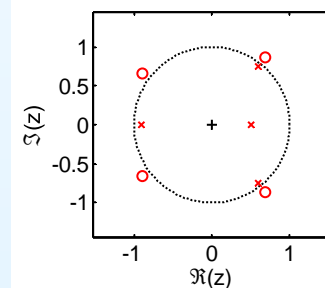
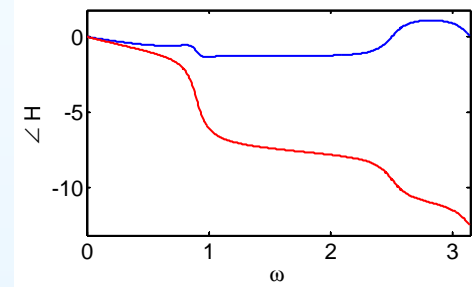
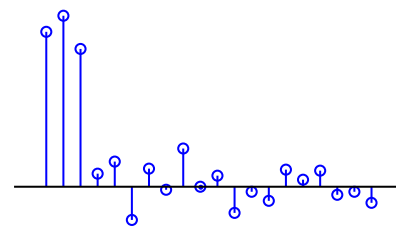
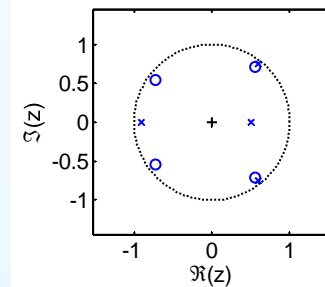
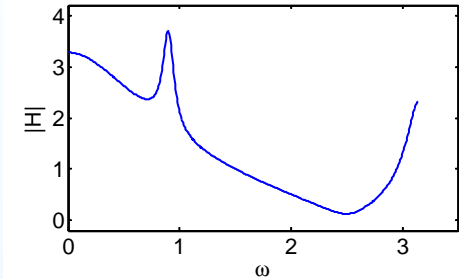
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$$2H(e^{j\omega}) = \sum_0^M h[n]e^{-j\omega n} + \sum_0^M h[M - n]e^{-j\omega(M-n)}$$

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$h[n]$ anti-symmetric:

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The phase of a linear phase filter is: $\angle H(e^{j\omega}) = \theta_0 - \alpha\omega$

Equivalently constant group delay: $\tau_H = -\frac{d\angle H(e^{j\omega})}{d\omega} = \alpha$

A filter has linear phase iff $h[n]$ is **symmetric** or **antisymmetric**:

$$h[n] = h[M - n] \quad \forall n \text{ or else } h[n] = -h[M - n] \quad \forall n$$

M can be even ($\Rightarrow \exists$ mid point) or odd ($\Rightarrow \nexists$ mid point)

Proof \Leftarrow :

$$\begin{aligned} 2H(e^{j\omega}) &= \sum_0^M h[n]e^{-j\omega n} + \sum_0^M h[M - n]e^{-j\omega(M-n)} \\ &= e^{-j\omega \frac{M}{2}} \sum_0^M h[n]e^{-j\omega(n - \frac{M}{2})} + h[M - n]e^{j\omega(n - \frac{M}{2})} \end{aligned}$$

$h[n]$ symmetric:

$$2H(e^{j\omega}) = 2e^{-j\omega \frac{M}{2}} \sum_0^M h[n] \cos\left(n - \frac{M}{2}\right) \omega$$

$h[n]$ anti-symmetric:

$$\begin{aligned} 2H(e^{j\omega}) &= -2je^{-j\omega \frac{M}{2}} \sum_0^M h[n] \sin\left(n - \frac{M}{2}\right) \omega \\ &= 2e^{-j\left(\frac{\pi}{2} + \omega \frac{M}{2}\right)} \sum_0^M h[n] \sin\left(n - \frac{M}{2}\right) \omega \end{aligned}$$

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- FIR Symmetries +
- IIR Frequency Response
- Negating z +
- Cubing z +
- Scaling z +
- Low-pass filter +
- Allpass filters +
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- Useful filters have difference equations:

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 - ▷ Pole bandwidth $\approx 2 |\log |p|| \approx 2 (1 - |p|)$

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For further details see Mitra: 6, 7.

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filter	filter a signal
impz	Impulse response
residuez	partial fraction expansion
grpdelay	Group Delay
freqz	Calculate filter frequency response