

9: Optimal IIR Design

- Error choices
- Linear Least Squares
- Frequency Sampling
- Iterative Solution
- Newton-Raphson
- Magnitude-only Specification
- Hilbert Relations
- Magnitude \leftrightarrow Phase Relation
- Summary
- MATLAB routines

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We want to find a filter $H(e^{j\omega}) = \frac{B(e^{j\omega})}{A(e^{j\omega})}$ that approximates a target response $D(\omega)$. Assume A is order N and B is order M .

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$$\text{Solution Error: } E_S(\omega) = W_S(\omega) \left(\frac{B(e^{j\omega})}{A(e^{j\omega})} - D(\omega) \right)$$

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Weight functions $W_*(\omega)$ are chosen to control relative errors at different frequencies. $W_S(\omega) = |D(\omega)|^{-1}$ gives constant dB error.

We actually want to minimize E_S but E_E is easier because it gives rise to linear equations.

However if $W_E(\omega) = \frac{W_S(\omega)}{|A(e^{j\omega})|}$, then $|E_E(\omega)| = |E_S(\omega)|$

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Differentiate with respect to \mathbf{x} :

$$d(\mathbf{e}^T \mathbf{e}) = d\mathbf{x}^T \mathbf{A}^T (\mathbf{Ax} - \mathbf{b}) + (\mathbf{x}^T \mathbf{A}^T - \mathbf{b}^T) \mathbf{A} d\mathbf{x}$$

[since $d(\mathbf{uv}) = d\mathbf{u} \mathbf{v} + \mathbf{u} d\mathbf{v}$]

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[since $d(\mathbf{uv}) = d\mathbf{u} \mathbf{v} + \mathbf{u} d\mathbf{v}$]
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This is zero for any $d\mathbf{x}$ iff $\mathbf{A}^T \mathbf{Ax} = \mathbf{A}^T \mathbf{b}$

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These are the **Normal Equations** (“Normal” because $\mathbf{A}^T \mathbf{e} = 0$)

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The **pseudoinverse** $\mathbf{x} = \mathbf{A}^+ \mathbf{b}$ works even if $\mathbf{A}^T \mathbf{A}$ is singular and finds the \mathbf{x} with minimum $\|\mathbf{x}\|^2$ that minimizes $\|\mathbf{e}\|^2$.

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This is a very widely used technique.

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For every ω we want: $0 = W(\omega) (B(e^{j\omega}) - D(\omega)A(e^{j\omega}))$

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$$\begin{aligned} \text{For every } \omega \text{ we want: } 0 &= W(\omega) (B(e^{j\omega}) - D(\omega)A(e^{j\omega})) \\ &= W(\omega) \left(\sum_{m=0}^M b[m]e^{-jm\omega} - D(\omega) \left(1 + \sum_{n=1}^N a[n]e^{-jn\omega} \right) \right) \end{aligned}$$

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$$\Rightarrow \begin{pmatrix} \mathbf{u}(\omega)^T & \mathbf{v}(\omega)^T \end{pmatrix} \begin{pmatrix} \mathbf{a} \\ \mathbf{b} \end{pmatrix} = W(\omega)D(\omega)$$

$$\text{where } \mathbf{u}(\omega)^T = -W(\omega)D(\omega) \begin{bmatrix} e^{-j\omega} & e^{-j2\omega} & \dots & e^{-jN\omega} \end{bmatrix}$$

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Least squares solution minimizes the E_E rather than E_S .

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- 4 Force $A(z)$ to be **stable**
Replace pole p_i by $(p_i^*)^{-1}$ whenever $|p_i| \geq 1$

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Replace pole p_i by $(p_i^*)^{-1}$ whenever $|p_i| \geq 1$
- 5 **Update weights:** $W_E(\omega_k) = \frac{W_S(\omega_k)}{|A(e^{j\omega_k})|}$

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But for faster convergence use Newton-Raphson . . .

Newton-Raphson

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Newton: To solve $f(x) = 0$ given an initial guess x_0 , we write

$$f(x) \approx f(x_0) + (x - x_0)f'(x_0) \Rightarrow x = x_0 - \frac{f(x_0)}{f'(x_0)}$$

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Converges very rapidly once x_0 is close to the solution

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So for each ω_k , we can write (omitting the ω and $e^{j\omega}$ arguments)

$$E_S \approx W_S \left(\frac{B_0}{A_0} - D \right) + \frac{W_S}{A_0} (B - B_0) - \frac{W_S B_0}{A_0^2} (A - A_0)$$

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$$\left(\begin{array}{cc} \frac{B_0}{DA_0} \mathbf{u}^T & \mathbf{v}^T \end{array} \right) \left(\begin{array}{c} \mathbf{a} \\ \mathbf{b} \end{array} \right) = W \left(A_0 D + \frac{B_0}{A_0} - B_0 \right)$$

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$$\text{where } W = \frac{W_S}{A_0}$$

Newton-Raphson

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Newton: To solve $f(x) = 0$ given an initial guess x_0 , we write

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Converges very rapidly once x_0 is close to the solution

So for each ω_k , we can write (omitting the ω and $e^{j\omega}$ arguments)

$$\begin{aligned} E_S &\approx W_S \left(\frac{B_0}{A_0} - D \right) + \frac{W_S}{A_0} (B - B_0) - \frac{W_S B_0}{A_0^2} (A - A_0) \\ &= \frac{W_S}{A_0} \left(B_0 - A_0 D + B - B_0 - \frac{B_0}{A_0} (A - 1) - \frac{B_0}{A_0} + B_0 \right) \end{aligned}$$

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for $n \in 1 : N$

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At each iteration, calculate $A_0(e^{j\omega_k})$ and $B_0(e^{j\omega_k})$ based on \mathbf{a} and \mathbf{b} from the previous iteration.

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At each iteration, calculate $A_0(e^{j\omega_k})$ and $B_0(e^{j\omega_k})$ based on \mathbf{a} and \mathbf{b} from the previous iteration.

Then use linear least squares to minimize the linearized E_S using the above equation replicated for each of the ω_k .

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If the filter specification only dictates the target magnitude: $|D(\omega)|$, we need to select the target phase.

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Solution:

Make an initial guess of the phase and then at each iteration

$$\text{update } \angle D(\omega) = \angle \frac{B(e^{j\omega})}{A(e^{j\omega})}.$$

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If $H(e^{j\omega})$ is **causal** and **minimum phase** then the magnitude and phase are not independent:

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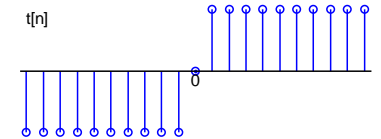
This result is a consequence of the **Hilbert Relations**.

Hilbert Relations

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We define $t[n] = u[n - 1] - u[-1 - n]$



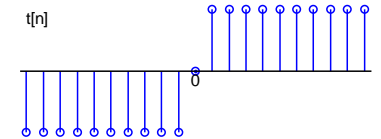
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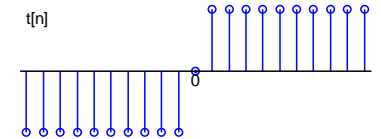
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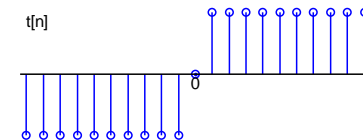
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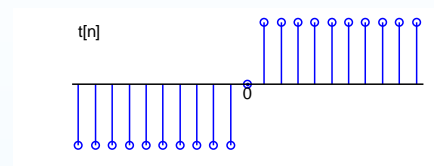
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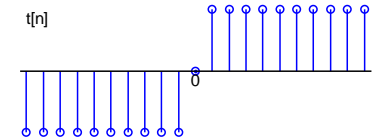
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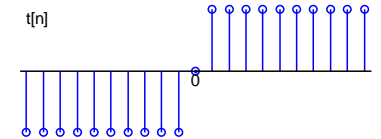
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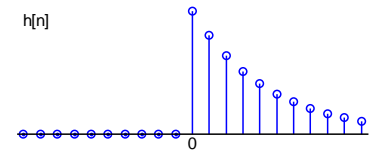
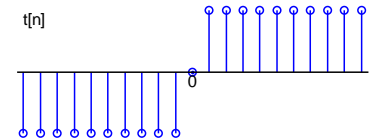
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$h[n]$



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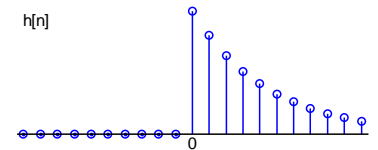
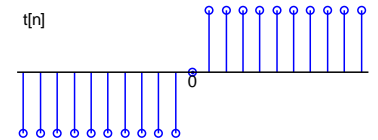
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Hilbert Relations

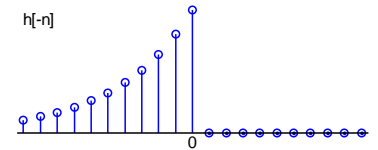
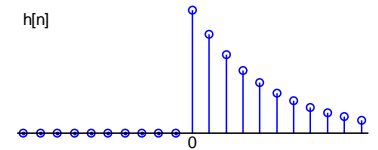
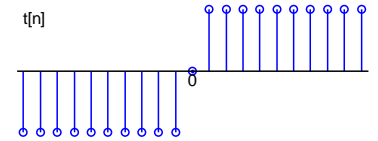
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Hilbert Relations

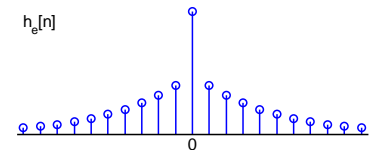
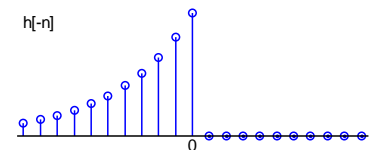
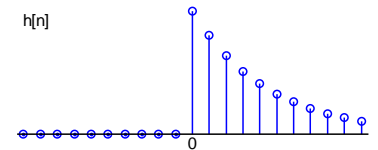
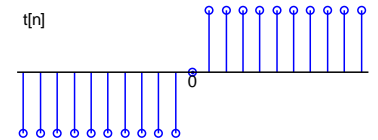
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$h[n] \rightarrow$ even/odd parts: $h_e[n] = \frac{1}{2} (h[n] + h[-n])$



Hilbert Relations

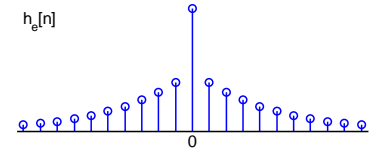
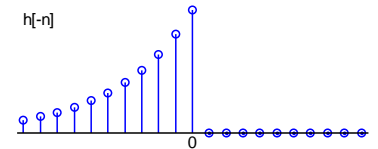
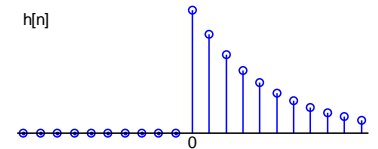
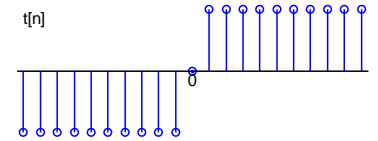
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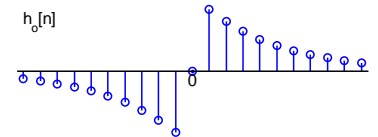
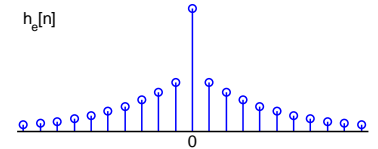
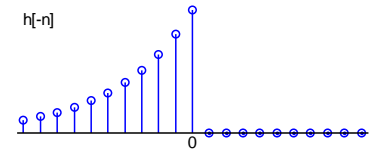
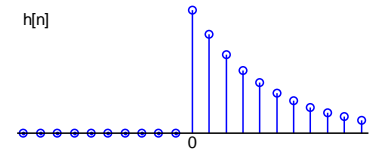
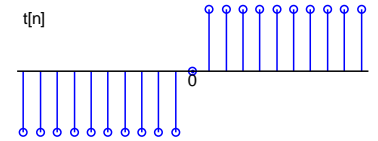
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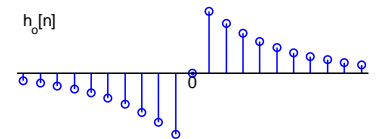
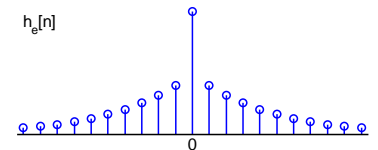
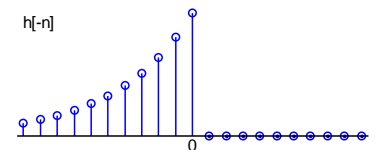
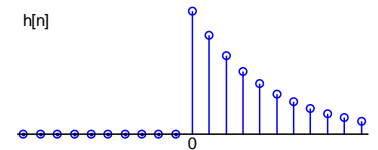
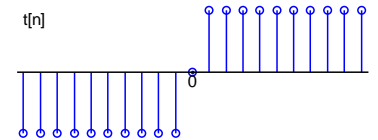
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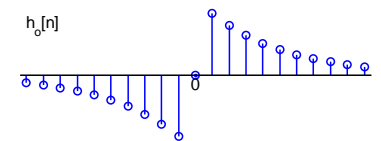
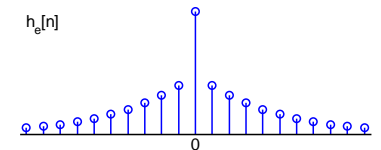
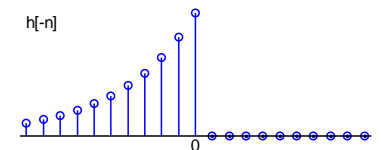
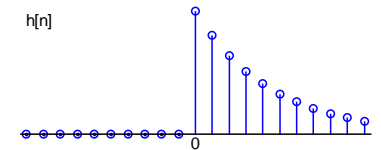
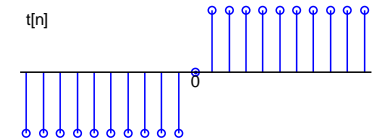
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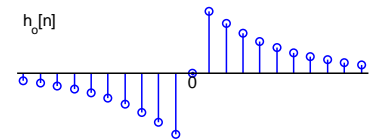
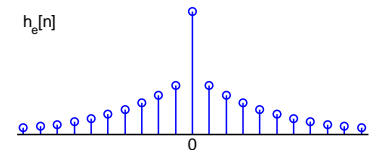
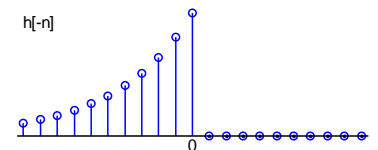
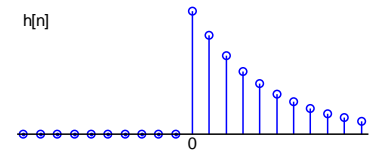
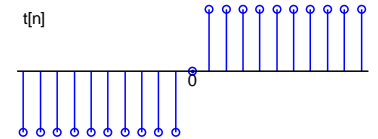
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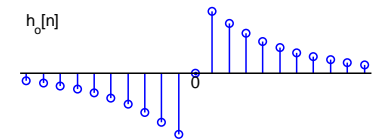
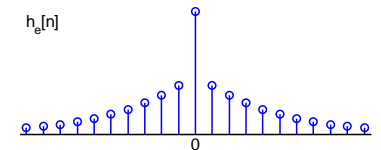
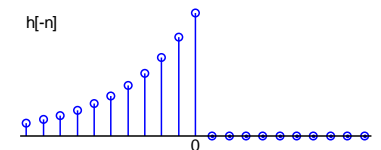
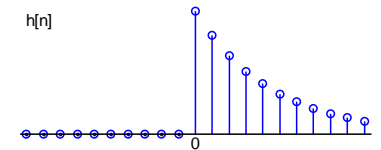
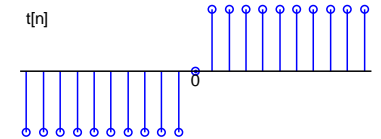
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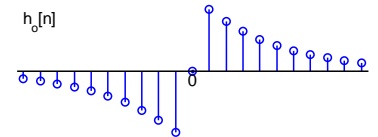
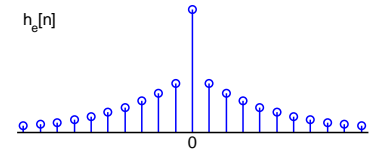
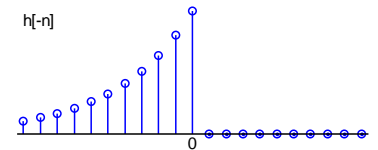
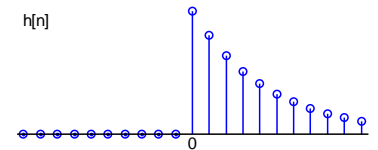
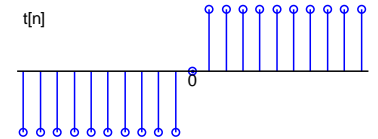
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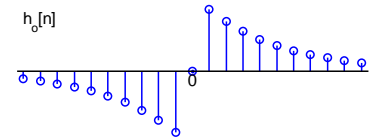
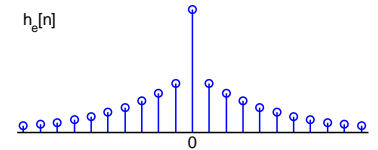
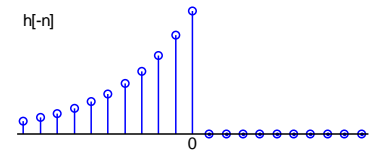
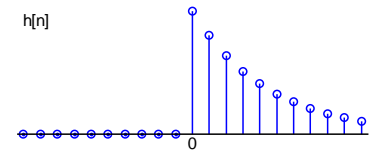
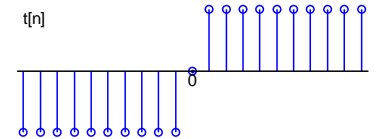
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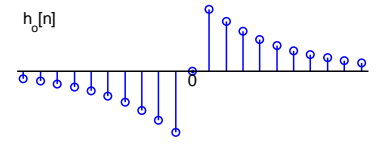
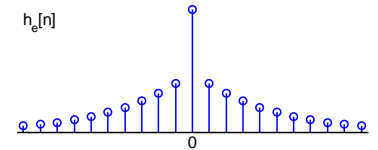
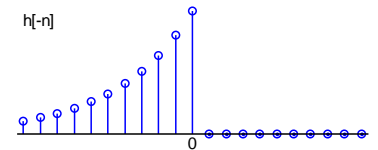
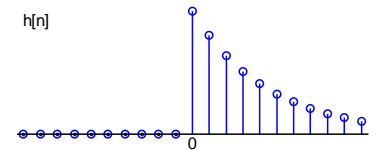
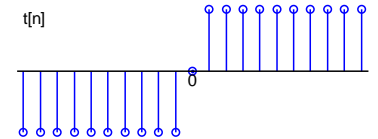
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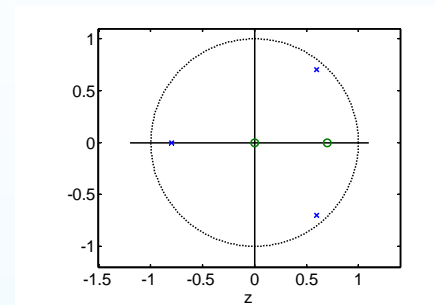
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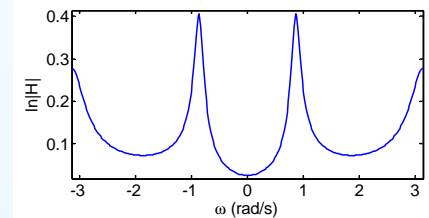
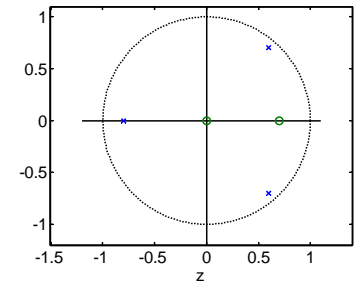
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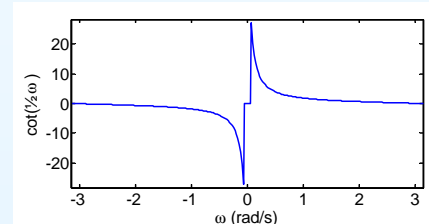
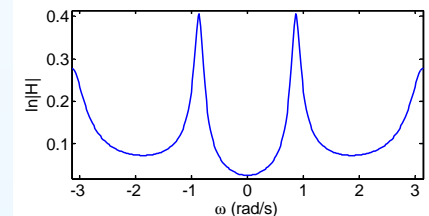
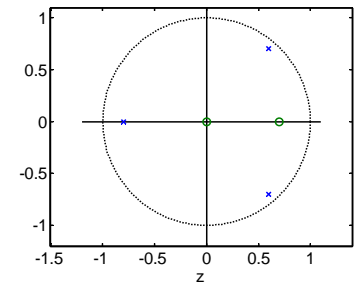
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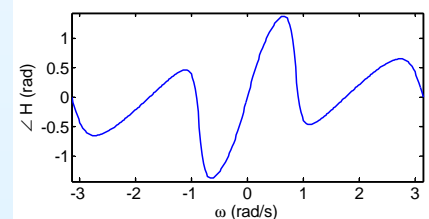
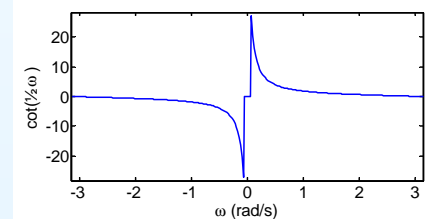
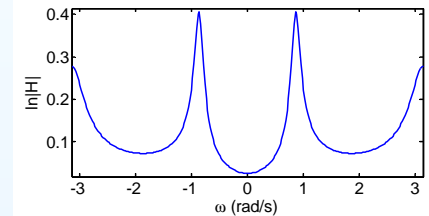
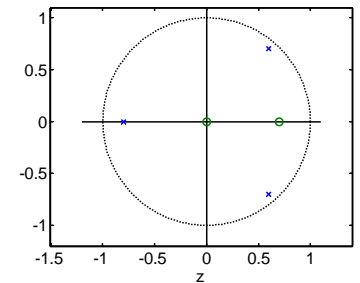
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For further details see Mitra: 9.

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invfreqz

IIR design for complex response