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- Halfband Filters
- Dyadic 1:8 Upsampler
- Rational Resampling
- Arbitrary Resampling +
- Polynomial Approximation

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Suppose we want to change the sample rate while preserving information: e.g. Audio $44.1 \text{ kHz} \leftrightarrow 48 \text{ kHz} \leftrightarrow 96 \text{ kHz}$

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LPF to new Nyquist bandwidth: $\omega_0 = \frac{\pi}{K}$



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 $\frac{x[n]}{LPF} - \frac{y[i]}{K:1}$

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Upsample:

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Rational ratio: $f_s \times \frac{P}{Q}$ LPF to lower of old and new Nyquist bandwidths: $\omega_0 = \frac{\pi}{\max(P,Q)}$



• Polyphase decomposition reduces computation by $K = \max(P, Q)$.

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- Fractional semi-Transition bandwidth, $\alpha = \frac{\Delta \omega}{2\omega_0}$, is typically fixed. e.g. $\alpha = 0.05 \implies M \approx \frac{dK}{7\pi\alpha} = 0.9dK$ (where $\omega_0 = \frac{\pi}{K}$)

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If K = 2 then the new Nyquist frequency is $\omega_0 = \frac{\pi}{2}$.



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If K=2 then the new Nyquist frequency is $\omega_0=\frac{\pi}{2}.$

We multiply ideal response $\frac{\sin \omega_0 n}{\pi n}$ by a Kaiser window.



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If $4 \mid M$ and we make the filter causal $(\times z^{-\frac{M}{2}})$, $H(z) = 0.5z^{-\frac{M}{2}} + z^{-1} \sum_{r=0}^{\frac{M}{2}-1} h_1[r] z^{-2r}$ where $h_1[r] = h[2r + 1 - \frac{M}{2}]$



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Half-band upsampler:

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Half-band upsampler:

We interchange the filters with the 1:2 block and use the commutator notation.



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Half-band upsampler:

We interchange the filters with the 1:2 block and use the commutator notation.

 $H_1(z)$ is symmetrical with $\frac{M}{2}$ coefficients so we need $\frac{M}{4}$ multipliers in total (input gain of 0.5 can usually be absorbed elsewhere).



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 $H_1(z)$ is symmetrical with $\frac{M}{2}$ coefficients so we need $\frac{M}{4}$ multipliers in total (input gain of 0.5 can usually be absorbed elsewhere).

Computation: $\frac{M}{4}$ multiplies per input sample

 $2H_1(z$

v|n

 $z^{-0.5M}$

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Upsample 1:2 $\rightarrow U(z)$:

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Upsample 1:2 $\rightarrow U(z)$: Filter $H_P(z)$ must remove image: $\Delta \omega = 0.2\pi$



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Suppose X(z): BW = $0.8\pi \Leftrightarrow \alpha = 0.2$

Upsample 1:2 $\rightarrow U(z)$: Filter $H_P(z)$ must remove image: $\Delta \omega = 0.2\pi$ For attenuation = 60 dB, $P \approx \frac{60}{3.5\Delta\omega} = 27.3$



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Upsample 1:2
$$\rightarrow V(z)$$
: $\Delta \omega = 0.6\pi \Rightarrow Q = 12$



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 $\begin{array}{c} x[i] & 0.125 \\ \hline @f_x & z^{-0.25P} & \downarrow u[j] \\ \hline & 2H_P(z) & 2H_Q(z) & \downarrow u[j] \\ \hline & 2H_R(z) & \downarrow u[n] \\ \hline & @8f \\ \hline & & 0.2 \\ \hline & & Upsample 1:2 \rightarrow U(z): \end{array}$

Filter $H_P(z)$ must remove image: $\Delta \omega = 0.2\pi$ For attenuation = 60 dB, $P \approx \frac{60}{3.5\Delta\omega} = 27.3$ Round up to a multiple of 4: P = 28

Upsample 1:2
$$\rightarrow V(z)$$
: $\Delta \omega = 0.6\pi \Rightarrow Q = 12$

Upsample 1:2
$$\rightarrow Y(z)$$
: $\Delta \omega = 0.8\pi \Rightarrow R = 8$



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Upsample 1:2 \rightarrow Y(z): $\Delta \omega = 0.8\pi \Rightarrow R = 8$ [diminishing returns + higher sample rate]



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Multiplication Count:

$$\left(1 + \frac{P}{4}\right) \times f_x + \frac{Q}{4} \times 2f_x + \frac{R}{4} \times 4f_x = 22f_x$$



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$$\left(1 + \frac{P}{4}\right) \times f_x + \frac{Q}{4} \times 2f_x + \frac{R}{4} \times 4f_x = 22f_x$$



Alternative approach using direct 1:8 upsampling: $\Delta \omega = 0.05\pi \Rightarrow M = 110 \Rightarrow 111 f_x$ multiplications (using polyphase)

Resampling: 13 - 4 / 10

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To resample by $\frac{P}{Q}$ do 1:*P* then LPF, then Q:1.



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x[n] **× × × × × × × ×**

v[s] odoodoodoodoodoodoodoodood

Resample by
$$\frac{P}{Q} \Rightarrow \omega_0 = \frac{\pi}{\max(P,Q)}$$

To resample by
$$\frac{P}{Q}$$
 do 1:*P* then LPF, then Q:1.

$$\frac{x[n]}{@f_x} 1:3 - H(z) \xrightarrow{v[s]} 5:1 \frac{y[i]}{@f_y}$$

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Resample by
$$\frac{P}{Q} \Rightarrow \omega_0 = \frac{\pi}{\max(P,Q)}$$

 $\Delta \omega \triangleq 2\alpha \omega_0 = \frac{2\alpha \pi}{\max(P,Q)}$

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$$H(z) = \sum_{p=0}^{P-1} z^{-p} H_p(z^P)$$

$$\frac{x[n]}{@1}H_p(z) \xrightarrow{v[s]} 5:1 \xrightarrow{v[i]}$$

$$h_0[r] \xrightarrow{\bullet}$$

$$h_1[r] \xrightarrow{\bullet}$$

$$h_2[r] \xrightarrow{\bullet}$$

$$r=0:R$$

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$$v[s]$$
 uses $H_p(z)$ with $p = s ext{ mod } P$

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v[s] uses $H_p(z)$ with $p = s \mod P$

Keep only every Q^{th} output:

To resample by $\frac{P}{Q}$ do 1:*P* then LPF, then Q:1. $\frac{x[n]}{@f_x}$ 1:3 $H(z) \frac{v[s]}{[0]}$ 5:1 $\frac{v[i]}{@f_y}$





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Resample by
$$\frac{P}{Q} \Rightarrow \omega_0 = \frac{\pi}{\max(P,Q)}$$

 $\Delta \omega \triangleq 2\alpha \omega_0 = \frac{2\alpha \pi}{\max(P,Q)}$

Polyphase: $H(z) = \sum_{p=0}^{P-1} z^{-p} H_p(z^P)$ Commutate coefficients:

v[s] uses $H_p(z)$ with $p = s \mod P$

Keep only every Q^{th} output: y[i] uses $H_p(z)$ with $p = Qi \mod P$ To resample by $\frac{P}{Q}$ do 1:*P* then LPF, then Q:1. $\frac{x[n]}{@f_x} 1:3 - H(z) \frac{v[s]}{[5:1]} 5:1 \frac{y[i]}{@f_y}$





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- Rational Resampling
- Arbitrary Resampling +
- Polynomial Approximation

+

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M+1 coeficients in all

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To resample by $\frac{P}{Q}$ do 1:*P* then LPF, then Q:1.

$$\frac{x[n]}{@f_x} 1:3 - H(z) \xrightarrow{v[s]} 5:1 \frac{y[i]}{@f_y}$$





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Multiplication rate:
$$\frac{2.7}{\alpha} \max\left(1, \frac{Q}{P}\right) \times f_y$$

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M+1 coeficients in all

DSP and Digital Filters (2017-10126)

Resampling: 13 - 5 / 10

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To resample by $\frac{P}{Q}$ do 1:*P* then LPF, then Q:1. $\frac{x[n]}{@f_{c}}$ 1:3 H(z) $\frac{v[s]}{[g_{c}]}$ 5:1 $\frac{y[i]}{@f_{c}}$





M+1 coeficients in all

Multiplication rate: $\frac{2.7}{\alpha} \max\left(1, \frac{Q}{P}\right) \times f_y = \frac{2.7}{\alpha} \max\left(f_y, f_x\right)$

Resampling: 13 - 5 / 10

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Sometimes need very large P and Q: e.g. $\frac{44.1 \text{ kHz}}{48 \text{ kHz}} = \frac{147}{160}$

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Alternatively, use any large integer Pand round down to the nearest sample: E.g. for y[i] at time $i\frac{Q}{P}$ use $h_p[r]$ where $p = (|iQ|)_{\text{mod }P}$



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Equivalent to converting to analog with zero-order hold and resampling at $f_y = \frac{P}{Q}$.



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Equivalent to converting to analog with zero-order hold and resampling at $f_y = \frac{P}{Q}$. Zero-order hold convolves with rectangular $\frac{1}{P}$ -wide window \Rightarrow multiplies

periodic spectrum by
$$\frac{\sin \frac{\Omega}{2P}}{\frac{\Omega}{2P}}$$
.

 $\frac{\Lambda I}{2P}$

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Alternatively, use any large integer Pand round down to the nearest sample: E.g. for y[i] at time $i\frac{Q}{P}$ use $h_p[r]$ where $p = (\lfloor iQ \rfloor)_{\text{mod } P}$

Equivalent to converting to analog with





zero-order hold and resampling at $f_y = \frac{P}{Q}$. Zero-order hold convolves with rectangular $\frac{1}{P}$ -wide window \Rightarrow multiplies periodic spectrum by $\frac{\sin \frac{\Omega}{2P}}{\frac{\Omega}{2P}}$. Resampling aliases Ω to $\Omega_{\text{mod } \frac{2P\pi}{Q}}$.

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Zero-order hold convolves with rectangular $\frac{1}{P}$ -wide window \Rightarrow multiplies periodic spectrum by $\frac{\sin \frac{\Omega}{2P}}{\frac{\Omega}{2P}}$. Resampling aliases Ω to $\Omega_{\text{mod } \frac{2P\pi}{Q}}$.

Unit power component at Ω_1 gives alias components with total power: $\sin^2 \frac{\Omega_1}{2P} \sum_{n=1}^{\infty} \left(\frac{2P}{2nP\pi + \Omega_1}\right)^2 + \left(\frac{2P}{2nP\pi - \Omega_1}\right)^2 \approx \frac{\omega_1^2}{4P^2} \frac{2\pi^2}{6\pi^2} = \frac{\Omega_1^2}{12P^2}$

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Equivalent to converting to analog with zero-order hold and resampling at $f_y = \frac{P}{Q}$.

Zero-order hold convolves with rectangular $\frac{1}{P}$ -wide window \Rightarrow multiplies periodic spectrum by $\frac{\sin \frac{\Omega}{2P}}{\frac{\Omega}{2P}}$. Resampling aliases Ω to $\Omega_{\text{mod } \frac{2P\pi}{Q}}$.

Unit power component at Ω_1 gives alias components with total power: $\sin^2 \frac{\Omega_1}{2P} \sum_{n=1}^{\infty} \left(\frac{2P}{2nP\pi + \Omega_1}\right)^2 + \left(\frac{2P}{2nP\pi - \Omega_1}\right)^2 \approx \frac{\omega_1^2}{4P^2} \frac{2\pi^2}{6\pi^2} = \frac{\Omega_1^2}{12P^2}$

For worst case, $\Omega_1 = \pi$, need P = 906 to get -60 dB \bigcirc

Resampling: 13 - 6 / 10

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Suppose P = 50 and H(z) has order M = 249H(z) is lowpass filter with $\omega_0 \approx \frac{\pi}{50}$



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 $\frac{x[n]}{a}$

 $h_0[r]$

 $h_{P-O}[r] \rightarrow @^P$

[i] uses h_{\parallel}

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Suppose P = 50 and H(z) has order M = 249H(z) is lowpass filter with $\omega_0 \approx \frac{\pi}{50}$ Split into 50 filters of length $R + 1 = \frac{M+1}{P} = 5$:





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Suppose P = 50 and H(z) has order M = 249H(z) is lowpass filter with $\omega_0 \approx \frac{\pi}{50}$ Split into 50 filters of length $R + 1 = \frac{M+1}{P} = 5$: $h_p[0]$ is the first P samples of h[m]





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Use a polynomial of order L to approximate each segment:

 $h_p[r] \approx f_r(\frac{p}{P})$ with $0 \leq \frac{p}{P} < 1$



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Use a polynomial of order *L* to approximate each segment:

 $h_p[r] \approx f_r(\frac{p}{P})$ with $0 \leq \frac{p}{P} < 1$

h[m] is smooth, so errors are low. E.g. error $< 10^{-3}$ for L = 4



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Resultant filter almost as good



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Use a polynomial of order L to approximate each segment:

 $h_p[r] \approx f_r(\frac{p}{D})$ with $0 \leq \frac{p}{D} < 1$

h[m] is smooth, so errors are low. E.g. error $< 10^{-3}$ for L = 4

- Resultant filter almost as good
- Instead of M + 1 = 250coefficients we only need (R+1)(L+1) = 25



(dB)

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Use a polynomial of order L to approximate each segment:

 $h_p[r] \approx f_r(\frac{p}{D})$ with $0 \leq \frac{p}{D} < 1$

h[m] is smooth, so errors are low. E.g. error $< 10^{-3}$ for L = 4

- Resultant filter almost as good
- Instead of M + 1 = 250coefficients we only need (R+1)(L+1) = 25where $R + 1 = \frac{2.7}{\alpha} \max\left(1, \frac{Q}{P}\right)$



(dB)

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$$\Delta[i] = i \frac{Q}{P} - n$$
 where $n = \left\lfloor i \frac{Q}{P} \right\rfloor$



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angle$

$$y[i] = \sum_{r=0}^{R} f_r(\Delta[i])x[n-r]$$

$$R + 1 = \frac{M+1}{P} = 5$$

$$\underbrace{x[n] \quad @1}_{H_{\Delta}(z)} \underbrace{y[i]}_{@^{P}/Q}$$

$$\underbrace{\Delta[i]}_{@^{P}/Q} \underbrace{[f_{0}(\Delta), \dots, f_{R}(\Delta)]}$$

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where $f_r(\Delta) = \sum_{l=0}^{L} b_l[r]\Delta^l$

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where $f_r(\Delta) = \sum_{l=0}^{L} b_l[r]\Delta^l$

$$y[i] = \sum_{r=0}^{R} \sum_{l=0}^{L} b_{l}[r] \Delta[i]^{l} x[n-r]$$

$$R + 1 = \frac{M+1}{P} = 5$$

$$\underbrace{x[n] \quad @ 1}_{H_{\Delta}(z)} \underbrace{y[i]}_{@^{P}/Q}$$

$$\underbrace{\Delta[i]}_{@^{P}/Q} [f_{0}(\Delta), \dots, f_{R}(\Delta)]$$

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- Resampling
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- Arbitrary Resampling +
- Polynomial Approximation

+

- Farrow Filter
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$$\Delta[i] = i \frac{Q}{P} - n \text{ where } n = \left\lfloor i \frac{Q}{P} \right\rfloor$$

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$R+1 = \frac{M+1}{P} = 5$
$x[n]$ @1 $H_{\Delta}(z)$ $y[i]$
$\frac{\Delta[i]}{a^{P}/a} [f_{0}(\Delta), \dots, f_{R}(\Delta)]$

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$R+1 = \frac{M}{R}$	$\frac{+1}{2} = 5$
x[n] @1	$H_{\Delta}(z) \xrightarrow{y[i]} y[i]$
$\frac{\Delta[i]}{a^{P}/a} [f_0(\Delta),$	$\dots, f_R(\Delta)$]

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Filter coefficients depend on fractional part of $i\frac{Q}{P}$:

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Horner's Rule: $y[i] = v_0[n] + \Delta (v_1[n] + \Delta (v_2[n] + \Delta (\cdots)))$




Farrow Filter

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Multiplication Rate:

Each $B_l(z)$ needs R + 1 per input sample Horner needs L per output sample





Farrow Filter

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Multiplication Rate:

Each $B_l(z)$ needs R + 1 per input sample Horner needs L per output sample

Total: $(L+1)(R+1)f_x + Lf_y = \frac{2.7(L+1)}{\alpha} \max\left(1, \frac{f_x}{f_y}\right)f_x + Lf_y$

 $R + 1 = \frac{M+1}{P} = 5$

 $\underline{(a)} H_{\Delta}(z)$

 $B_1(z)$ $v_1[n]$

 $B_2(z)$ $v_2[n]$

 $B_3(z)$ $v_3[n]$

 $B_L(z)$

 $v_L[n]$

 $\Delta[i]$

 $(a)^{P}$

Farrow Filter

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$$= \sum_{l=0}^{L} \Delta[i]^{l} \sum_{r=0}^{R} b_{l}[r]x[n-r]$$

$$= \sum_{l=0}^{L} \Delta[i]^{l}v_{l}[n]$$

where $v_{l}[n] = b_{l}[n] * x[n]$
[like a Taylor series expansion]

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 $y[i] = v_0[n] + \Delta \left(v_1[n] + \Delta \left(v_2[n] + \Delta \left(\cdots \right) \right) \right)$

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 $\begin{array}{c} \hline B_{0}(z) & v_{0}[n] & v_{1}[i] \\ \hline B_{1}(z) & v_{1}[n] & \\ \hline B_{2}(z) & v_{2}[n] & \\ \hline B_{3}(z) & v_{3}[n] & \\ \hline B_{L}(z) & v_{L}[n] \\ \hline \Delta[i] & P_{L} \end{array}$

 $R + 1 = \frac{M+1}{P} = 5$

 $\underline{(a)} H_{\Delta}(z)$

$$1 \approx \frac{2.7}{\alpha} \max\left(1, \frac{Q}{P}\right)$$

R+

$$+$$

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- Transition band centre at ω_0
 - \circ ω_0 = the lower of the old and new Nyquist frequencies
 - Transition width = $\Delta \omega = 2 \alpha \omega_0$, typically $\alpha \approx 0.1$

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- Rational resampling $\times \frac{P}{O}$
 - # multiplies per second: $\frac{2.7}{\alpha} \max(f_y, f_x)$
 - # coefficients: $\frac{2.7}{\alpha} \max{(\tilde{P}, Q)}$

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 - approximate filter impulse response with polynomial segments
 - arbitrary, time-varying, resampling ratios
 - # multiplies per second: $\frac{2.7(L+1)}{\alpha} \max(f_y, f_x) \times \frac{f_x}{f_y} + Lf_y$
 - $\triangleright \quad \approx (L+1) \, rac{f_x}{f_y}$ times rational resampling case
 - # coefficients: $\frac{2.7}{\alpha} \max(P, Q) \times \frac{L+1}{P}$
 - \circ coefficients are independent of f_y when upsampling

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For further details see Mitra: 13 and Harris: 7, 8.

MATLAB routines

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gcd(p,q)	Find $\alpha p + \beta q = 1$ for coprime p , q
polyfit	Fit a polynomial to data
polyval	Evaluate a polynomial
upfirdn	Perform polyphase filtering
resample	Perform polyphase resampling